Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences with the TMS320 DSP Family

APPLICATION REPORT: SPRA291

Robert Matusiak
Member, Group Technical Staff

Digital Signal Processing Solutions
December 1997
IMPORTANT NOTICE

Texas Instruments (TI) reserves the right to make changes to its products or to discontinue any semiconductor product or service without notice, and advises its customers to obtain the latest version of relevant information to verify, before placing orders, that the information being relied on is current.

TI warrants performance of its semiconductor products and related software to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are utilized to the extent TI deems necessary to support this warranty. Specific testing of all parameters of each device is not necessarily performed, except those mandated by government requirements.

Certain application using semiconductor products may involve potential risks of death, personal injury, or severe property or environmental damage (“Critical Applications”).

TI SEMICONDUCTOR PRODUCTS ARE NOT DESIGNED, INTENDED, AUTHORIZED, OR WARRANTED TO BE SUITABLE FOR USE IN LIFE-SUPPORT APPLICATIONS, DEVICES OR SYSTEMS OR OTHER CRITICAL APPLICATIONS.

Inclusion of TI products in such applications is understood to be fully at the risk of the customer. Use of TI products in such applications requires the written approval of an appropriate TI officer. Questions concerning potential risk applications should be directed to TI through a local SC sales office.

In order to minimize risks associated with the customer’s applications, adequate design and operating safeguards should be provided by the customer to minimize inherent or procedural hazards.

TI assumes no liability for applications assistance, customer product design, software performance, or infringement of patents or services described herein. Nor does TI warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right of TI covering or relating to any combination, machine, or process in which such semiconductor products or services might be or are used.

Copyright © 1997, Texas Instruments Incorporated
TRADEMARKS

TI is a trademark of Texas Instruments Incorporated.

Other brands and names are the property of their respective owners.
## CONTACT INFORMATION

<table>
<thead>
<tr>
<th>Service</th>
<th>Phone Number</th>
</tr>
</thead>
<tbody>
<tr>
<td>US TMS320 HOTLINE</td>
<td>(281) 274-2320</td>
</tr>
<tr>
<td>US TMS320 FAX</td>
<td>(281) 274-2324</td>
</tr>
<tr>
<td>US TMS320 BBS</td>
<td>(281) 274-2323</td>
</tr>
<tr>
<td>US TMS320 email</td>
<td><a href="mailto:dsph@ti.com">dsph@ti.com</a></td>
</tr>
</tbody>
</table>
**Contents**

**Abstract** .............................................................................................................................................. 7

**Product Support** .................................................................................................................................. 8
  Related Documentation .......................................................................................................................... 8
  World Wide Web .................................................................................................................................... 8

**Introduction** ........................................................................................................................................... 9

**Basics of the DFT and FFT** .................................................................................................................. 10

**Efficient Computation of the DFT of Real Sequences** ......................................................................... 12
  Efficient Computation of the DFT of Two Real Sequences ..................................................................... 12
  Efficient Computation of the DFT of a \(2N\)-Point Real Sequence ......................................................... 15

**TMS320C62xx Architecture and Tools Overview** ................................................................................. 20

**Implementation and Optimization of Real-Valued DFTs** ................................................................. 26

**Summary** ............................................................................................................................................... 29

**Appendix A. Derivation of Equations Used to Compute the DFT/IDFT of Two Real Sequences** ............ 30
  Forward Transform ............................................................................................................................... 30
  Inverse Transform ............................................................................................................................... 33

**Appendix B. Derivation of Equations Used to Compute the DFT/IDFT of a Real Sequence** .................... 34
  Forward Transform ............................................................................................................................... 34
  Inverse Transform ............................................................................................................................... 36

**Appendix C. C Implementations of the DFT of Real Sequences** .......................................................... 41
  Implementation Notes ........................................................................................................................... 41
  Implementation Notes ........................................................................................................................... 62

**Appendix D. Optimized C Callable 'C62xx Assembly Language Functions Used to Implement the DFT of Real Sequences** ........................................................................................................ 74
  Implementation Notes ........................................................................................................................... 74

**References** ............................................................................................................................................. 94
Figures
Figure 1. TMS320C6201 DSP Block Diagram ............................................................... 20
Figure 2. Code Development Flow Chart................................................................. 23

Examples
Example 1. Efficient Computation of the DFT of a 2N-Point Real Sequence .......... 26
Example 2. Efficient Computation of the DFT of Two Real Sequences................ 26
Example 3. Efficient Computation of the DFT of a 2N-Point Real Sequence ....... 27
Example 4. Efficient Computation of the DFT of Two Real Sequences............... 27
Example 5. Efficient Computation of the DFT of a 2N-Point Real Sequence ....... 28
Example 6. Efficient Computation of the DFT of Two Real Sequences.............. 28
Example 7. realdft1.c File .......................................................................................... 43
Example 8. split1.c File ............................................................................................ 47
Example 9. data1.c File ............................................................................................ 48
Example 10. params1.h File ...................................................................................... 49
Example 11. realdft2.c File ....................................................................................... 50
Example 12. split2.c File .......................................................................................... 53
Example 13. data2.c File .......................................................................................... 55
Example 14. params2.h File ...................................................................................... 56
Example 15. dft.c File ............................................................................................... 57
Example 16. params.h File ....................................................................................... 59
Example 17. vectors.asm ......................................................................................... 60
Example 18. lnk.cmd ............................................................................................... 61
Example 19. realdft3.c File ....................................................................................... 63
Example 20. realdft4.c File ....................................................................................... 67
Example 21. radix4.c File ......................................................................................... 70
Example 22. digit.c File ........................................................................................... 71
Example 23. digitgen.c File ...................................................................................... 72
Example 24. splitgen.c File ...................................................................................... 73
Example 25. split1.asm File ..................................................................................... 75
Example 26. split2.asm File ..................................................................................... 81
Example 27. radix4.asm File .................................................................................... 86
Example 28. digit.asm File ...................................................................................... 91

Tables
Table 1. Comparison of Computational Complexity for Direct Computation of the DFT
vs. the Radix-2 FFT Algorithm .............................................................................. 11
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences with the TMS320 DSP Family

Abstract

The Fast Fourier Transform (FFT) is an efficient computation of the Discrete Fourier Transform (DFT) and one of the most important tools used in digital signal processing applications. Because of its well-structured form, the FFT is a benchmark in assessing digital signal processor (DSP) performance.

The development of FFT algorithms has assumed an input sequence consisting of complex numbers. This is because complex phase factors, or twiddle factors, result in complex variables. Thus, FFT algorithms are designed to perform complex multiplications and additions. However, the input sequence consists of real numbers in a large number of real applications.

This application report discusses the theory and usage of two algorithms used to efficiently compute the DFT of real valued sequences as implemented on the Texas Instruments (TI™) TMS320C6x (‘C6x).

The first algorithm performs the DFT of two N-point real-valued sequences using one N-point complex DFT and additional computations.

The second algorithm performs the DFT of a 2N-point real-valued sequence using one N-point complex DFT and additional computations. Implementations of these additional computations, referred to as the split operation, are presented both in C and ‘C6x assembly language. For implementation on the ‘C6x, optimization techniques in both C and assembly are covered.
Product Support

Related Documentation

TMS320C62xx Programmers Guide, literature number SPRU198

TMS320C6x Optimizing C Compiler User’s Guide, literature number SPRU187A

TMS320C6x Assembly Language Tools User’s Guide, literature number SPRU186

TMS320C62x C Source Debugger User’s Guide, literature number SPRU188D

World Wide Web

Our World Wide Web site at www.ti.com contains the most up-to-date product information, revisions, and additions. Users registering with TI&ME can build custom information pages and receive new product updates automatically via email.


**Introduction**

TI’s ‘C6x family of high performance, fixed-point DSPs provide architectural and speed improvements that make the FFT computation faster and easier to program than other fixed-point DSPs.

The ‘C6x family devices are based on an advanced VLIW (Very Long Instruction Word) CPU with eight functional units that include two multipliers and six arithmetic logic units (ALUs). The CPU can execute up to eight instructions per cycle. Complementing the architecture is a very efficient C compiler that increases performance and reduces code development time. The ‘C6x architecture and development tools feature are discussed in this application report along with the following topics:

- Theory of DFTs of real-valued sequences
- Algorithm implementation
- ‘C6x CPU features
- ‘C6x development tools
- Optimizing C code for the ‘C6x
- C callable assembly language functions for the ‘C6x
Basics of the DFT and FFT

Methods of performing the DFT of real sequences involve complex-valued DFTs. This section reviews the basics of the DFT and FFT.

The DFT is viewed as a frequency domain representation of the discrete-time sequence $x(n)$. The $N$-point DFT of finite-duration sequence $x(n)$ is defined as

$$X(k) = \sum_{n=0}^{N-1} x(n)W_N^{kn} \quad k = 0,1,\ldots,N-1$$

(1)

and the inverse DFT (IDFT) is defined as

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k)W_N^{-kn} \quad n = 0,1,\ldots,N-1$$

(2)

where

$$W_N^{nk} = e^{-j2\pi nk/N}$$

The $W_N^{kn}$ factor is also referred to as the twiddle factor.

Observation of the above equations shows that the computational requirements of the DFT increase rapidly as the number of samples in the sequence $N$ increases. Because of the large computational requirements, direct implementation of the DFT of large sequences has not been practical for real-time applications. However, the development of fast algorithms known as FFTs has made implementation of DFT practical in real-time applications.

The definition of FFT is the same as DFT but the method of computation differs. The basics of FFT algorithms involve a divide-and-conquer approach in which an $N$-point DFT is divided into successively smaller DFTs. Many FFT algorithms have been developed, such as radix-2, radix-4, and mixed radix; in-place and not-in-place; and decimation-in-time and decimation-in-frequency.

In most FFT algorithms, restrictions may apply. For example, a radix-2 FFT restricts the number of samples in the sequence to a power of two.

In addition, some FFT algorithms require the input or output to be re-ordered. For example, the radix-2 decimation-in-frequency algorithm requires the output to be bit-reversed. It is up to implementers to choose the FFT algorithm that best fits their application.
Table 1 compares the number of math computations involved in

direct computation of the DFT versus the radix-2 FFT algorithm.
As you can see, the speed improvement of the FFT increases as

$N$ increases. Detailed descriptions of the DFT and FFT can be

found in references.\textsuperscript{123} The following sections describe methods of

efficiently computing the DFT of real-valued sequences using

complex-valued DFTs/IDFTs.

\textbf{Table 1. Comparison of Computational Complexity for Direct Computation of the

DFT vs. the Radix-2 FFT Algorithm}

<table>
<thead>
<tr>
<th>Number of Points</th>
<th>Direct Computation of the DFT</th>
<th>Radix-2 FFT</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Complex Multiples</td>
<td>Complex Additions</td>
</tr>
<tr>
<td>$N$</td>
<td>$N^2$</td>
<td>$N^2-N$</td>
</tr>
<tr>
<td>4</td>
<td>16</td>
<td>12</td>
</tr>
<tr>
<td>16</td>
<td>256</td>
<td>240</td>
</tr>
<tr>
<td>64</td>
<td>4096</td>
<td>4032</td>
</tr>
<tr>
<td>256</td>
<td>65536</td>
<td>65280</td>
</tr>
<tr>
<td>1024</td>
<td>1048576</td>
<td>1047552</td>
</tr>
</tbody>
</table>
Efficient Computation of the DFT of Real Sequences

In many real applications, the data sequences to be processed are real valued. Even though the data is real, complex-valued DFT algorithms can still be used. One simple approach creates a complex sequence from the real sequence; that is, real data for the real components and zeros for the imaginary components. The complex DFT can then be applied directly. However, this method is not efficient. This section shows you how to use the complex-valued DFT algorithms to efficiently process real-valued sequences.

Efficient Computation of the DFT of Two Real Sequences

Suppose \( x_1(n) \) and \( x_2(n) \) are real-valued sequences of length \( N \), and \( x(n) \) is a complex-valued sequence defined as

\[
x(n) = x_1(n) + jx_2(n) \quad 0 \leq n \leq N - 1
\]

The DFT of the two \( N \)-length sequences \( x_1(n) \) and \( x_2(n) \) can be found by performing a single \( N \)-length DFT on the complex-valued sequence and some additional computation. These additional computations are referred to as the split operation and shown below.

\[
X_1(k) = \frac{1}{2} \left[ X(k) + X^{*}(N - k) \right] \quad k = 0, 1, \ldots, N - 1
\]

\[
X_2(k) = \frac{1}{2} j \left[ X(k) - X^{*}(N - k) \right]
\]

As you can see from the above equations, the transforms of \( x_1(n) \) and \( x_2(n) \), \( X_1(k) \) and \( X_2(k) \) respectively, are solved by computing one complex-valued DFT, \( X(k) \), and some additional computations.

Now assume we want to get back \( x_1(n) \) and \( x_2(n) \) from \( X_1(k) \) and \( X_2(k) \), respectively. As with the forward DFT, the IDFT of \( X_1(k) \) and \( X_2(k) \) is found using a single complex-valued DFT. Because the DFT operation is linear, the DFT of equation (4) can be expressed as

\[
X(k) = X_1(k) + jX_2(k)
\]

This shows that \( X(k) \) can be expressed in terms of \( X_1(k) \) and \( X_2(k) \); thus, taking the inverse DFT of \( X(k) \), we get \( x(n) \), which gives us \( x_1(n) \) and \( x_2(n) \).
The above equations require complex arithmetic not directly supported by DSPs; thus, to implement these complex-valued equations, it is helpful to express the real and imaginary terms in real arithmetic. The forward DFT of the equations shown in (5) can be written as follows:

\[
X_1 r(k) = \frac{1}{2} [Xr(k) + Xr(N - k)] \quad \text{and} \quad X_1 i(k) = \frac{1}{2} [Xi(k) - Xi(N - k)] \\
k = 0, 1, \ldots, N - 1
\]

\[
X_2 r(k) = \frac{1}{2} [Xi(k) + Xi(N - k)] \quad \text{and} \quad X_2 i(k) = -\frac{1}{2} [Xr(k) - Xr(N - k)]
\]

In addition, because the DFT of real-valued sequences has the properties of complex conjugate symmetry and periodicity, the number of computations in (7) can be reduced. Using the properties, the equations in (7) can be re-written as follows:

\[
X_1 r(0) = Xr(0) \quad \quad X_1 i(0) = 0 \\
X_2 r(0) = Xi(0) \quad \quad X_2 i(0) = 0 \\
X_1 r(N / 2) = Xr(N / 2) \quad \quad X_1 i(N / 2) = 0 \\
X_2 r(N / 2) = Xi(N / 2) \quad \quad X_2 i(N / 2) = 0
\]

\[
X_1 r(k) = \frac{1}{2} [Xr(k) + Xr(N - k)] \quad X_1 i(k) = \frac{1}{2} [Xi(k) - Xi(N - k)] \\
X_2 r(k) = \frac{1}{2} [Xi(k) + Xi(N - k)] \quad X_2 i(k) = -\frac{1}{2} [Xr(k) - Xr(N - k)] \\
X_1 r(N - k) = X_1 r(k) \quad X_1 i(N - k) = -X_1 i(k) \\
X_2 r(N - k) = X_2 r(k) \quad X_2 i(N - k) = -X_2 i(k) \\
k = 1, 2, \ldots, N / 2 - 1
\]

Similarly, the additional computation involved in computing the IDFT can be written as follows:

\[
Xr(k) = X_1 r(k) - X_2 i(k) \quad \quad k = 0, 1, \ldots, N - 1
\]

\[
Xi(k) = X_1 i(k) + X_2 r(k)
\]
See Appendix A for a detailed derivation of these equations.

Now that we have the equations for the split operation used in computing the DFT of two real-valued sequences, we turn to the following steps, which outline how to use the equations. The forward DFT is outlined as follows.

**Step 1:** Form the $N$-point complex-valued sequence $x(n)$ from the two $N$-length sequences $x_1(n)$ and $x_2(n)$.

for $n=0, \ldots, N-1$

\[
\begin{align*}
x_r(n) &= x_1(n) \\
x_i(n) &= x_2(n)
\end{align*}
\]

**Step 2:** Compute the $N$-length complex DFT of $x(n)$.

\[X(k) = \text{DFT}[x(n)]\]

**NOTE:**
The DFT can be any efficient DFT algorithm (such as one of the various FFT algorithms) but the output must be in normal order.

**Step 3:** Compute the split operation equations.

\[
\begin{align*}
X_1r(0) &= Xr(0) & X_1i(0) &= 0 \\
X_2r(0) &= Xi(0) & X_2i(0) &= 0 \\
X_1r(N/2) &= Xr(N/2) & X_1i(N/2) &= 0 \\
X_2r(N/2) &= Xi(N/2) & X_2i(N/2) &= 0
\end{align*}
\]

for $k=1, \ldots, N/2-1$

\[
\begin{align*}
X_1r(k) &= 0.5 \times [Xr(k) + Xr(N-k)] & X_1i(k) &= 0.5 \times [Xi(k) - Xi(N-k)] \\
X_2r(k) &= 0.5 \times [Xi(k) + Xi(N-k)] & X_2i(k) &= -0.5 \times [Xr(k) - Xr(N-k)] \\
X_1r(N-k) &= X_1r(k) & X_1i(N-k) &= -X_1i(k) \\
X_2r(N-k) &= X_2r(k) & X_2i(N-k) &= -X_2i(k)
\end{align*}
\]

For two frequency domain sequences, $X_1(k)$ and $X_2(k)$, derived from real-valued sequences, perform the following steps to take the IDFT of $X_1(k)$ and $X_2(k)$.

**Step 1:** Form a single complex-valued sequence $X(k)$ from $X_1(k)$ and $X_2(k)$ using the IDFT split equations.

for $k=0, \ldots, N-1$

\[
\begin{align*}
x_r(k) &= X_1r(k) - X_2i(k) \\
x_i(k) &= X_1i(k) + X_2r(k)
\end{align*}
\]

**Step 2:** Compute the $N$-length IDFT of $X(k)$.

\[x(n) = \text{IDFT}[X(k)]\]
As with the forward DFT, the IDFT can be any efficient IDFT algorithm. The IDFT can be computed using the forward DFT and some conjugate operations.

\[ x(n) = [\text{DFT} \{X^*(k)\}]^* \]

where:

* is the complex conjugate operator

**Step 3:** From \(x(n)\), form \(x_1(n)\) and \(x_2(n)\).

for \(n = 0, 1, \ldots, N-1\)

\[ x_1(n) = x_r(n) \]

\[ x_2(n) = x_i(n) \]

Appendix C contains C implementation of the outlined DFT and IDFT algorithms.

**Efficient Computation of the DFT of a \(2N\)-Point Real Sequence**

Assume \(g(n)\) is a real-valued sequence of \(2N\) points. We outline the equations involved in obtaining the \(2N\)-point DFT of \(g(n)\) from the computation of one \(N\)-point complex-valued DFT. First, we subdivide the \(2N\)-point real sequence into two \(N\)-point sequences as follows:

\[ x_1(n) = g(2n) \quad 0 \leq n \leq N - 1 \quad (10) \]

\[ x_2(n) = g(2n + 1) \]

And define \(x(n)\) to be the \(N\)-point complex-valued sequence:

\[ x(n) = x_1(n) + jx_2(n) \quad 0 \leq n \leq N - 1 \quad (11) \]

The DFT of \(g(n)\), \(G(k)\), can be computed using

\[ G(k) = X(k)A(k) + X^*(N - k)B(k) \quad k = 0, 1, \ldots, N - 1 \quad (12) \]

with \(X(N) = X(0)\)

where

\[ A(k) = \frac{1}{2}(1 - jW_{2N}^k) \quad \text{and} \quad B(k) = \frac{1}{2}(1 + jW_{2N}^k) \quad (13) \]

As you can see, we have computed the DFT of a \(2N\)-point sequence from one \(N\)-point DFT and additional computations, which we call the split operation.
Similarly, if we have a frequency domain $2N$-point sequence, which was derived from a real-valued sequence, we can use an $N$-point IDFT to obtain the time domain $2N$-point real-valued sequence using the following equation:

$$X(k) = G(k)A^*(k) + G^*(N - k)B^*(k) \quad k = 0, 1, \ldots, N - 1$$

with $G(N) = G(0)$

The equations shown in (12) and (14) are of the same form. Equation (14) can be obtained from equation (12) if $G(k)$ is swapped with $X(k)$, and $A(k)$ and $B(k)$ are complex conjugated. Thus, equations (12) and (14) can be implemented with one common split function.

**NOTE:**

In implementing these equations, $A(k)$, $A^*(k)$, $B(k)$, and $B^*(k)$ can be pre-computed and stored in a table. Their values can thus be obtained by table look-up as opposed to arithmetic computation. The result is a large computational savings because the sine and cosine functions required by twiddle factors do not need to be computed when performing the split. (A detailed derivation of these equations is provided in Appendix B.)

As in the previous section, Efficient Computation of the DFT of Two Real Sequences, when implementing the above equations, it is useful to express them in their real and imaginary terms.

Only $N$ points of $G(k)$ are computed in equation (12) because other $N$ points can be found using the complex conjugate property. This is applied to the following equation.

$$Gr(k) = Xr(k)Ar(k) - Xi(k)Ai(k) + Xr(N - k)Br(k) + Xi(N - k)Bi(k)$$

$$Gi(k) = Xi(k)Ar(k) + Xr(k)Ai(k) + Xr(N - k)Bi(k) - Xi(N - k)Br(k)$$

$$Gr(2N - k) = Gr(k) \quad k = 0, 1, \ldots, N - 1$$

$$Gi(2N - k) = -Gi(k) \quad X(N) = X(0)$$

$$Gr(N) = Gr(0) - Gi(0)$$

$$Gi(N) = 0$$

As with the forward DFT, the equations for the IDFT can be expressed in their real and imaginary terms as follows.
\[ X_r(k) = G_r(k)A_r(k) - G_i(k)(-A_i(k)) + G_r(N-k)B_r(k) + G_i(N-k)(B_i(k)) \]
\[ X_i(k) = G_i(k)A_r(k) + G_r(k)(-A_i(k)) + G_r(N-k)(-B_i(k)) - G_i(n-k)B_r(k) \]

Now that we have the equations for the split operation to compute the DFT of a real-valued sequence, the steps for using these equations are outlined. The forward DFT is outlined first.

**Step 1:** Initialize \(A(k)'s\) and \(B(k)'s\).

Real applications usually perform this only once during a power-up or initialization sequence. These values can be pre-stored in a boot ROM or computed. In either case, once they are generated, this step is no longer needed when performing the DFT. The pseudo code for generating them is given below.

for \(k = 0, 1, \ldots, N-1\)

\[
\begin{align*}
A_i(k) &= -\cos(\pi k/N) \\
A_r(k) &= 1 - \sin(\pi k/N) \\
B_i(k) &= \cos(\pi k/N) \\
B_r(k) &= 1 + \sin(\pi k/N)
\end{align*}
\]

**Step 2:** Let \(g(n)\) be a \(2N\)-point real sequence. From \(g(n)\), form the \(N\)-point complex valued sequence

\[ x(n) = x_1(n) + jx_2(n) \]

where

\[ x_1(n) = g(2n) \]
\[ x_2(n) = g(2n + 1) \]

for \(n = 0, 1, \ldots, N-1\)

\[ x_r(n) = g(2n) \]
\[ x_i(n) = g(2n + 1) \]

**Step 3:** Perform an \(N\)-point complex FFT on the complex valued sequence \(x(n)\).

\[ X(k) = \text{DFT}[x(n)] \]

**NOTE:**

The FFT can be any DFT method, such as radix-2, radix-4, mixed radix, direct implementation of the DFT, etc. However, the DFT output must be in normal order.
Step 4: Implement the split operation equations.

\[ X(N) = X(0) \]
\[ Gr(N) = Gr(0) - Gi(0) \]
\[ Gi(N) = 0 \]

for \( k = 0, 1, \ldots, N-1 \)

\[ Gr(k) = Xr(k)Ar(k) - Xi(k)Ai(k) + Xr(N-k)Br(k) + Xi(N-k)Bi(k) \]
\[ Gi(k) = Xi(k)Ar(k) + Xr(k)Ai(k) + Xr(N-k)Bi(k) - Xi(N-k)Br(k) \]
\[ Gr(2N-k) = Gr(k) \]
\[ Gi(2N-k) = -Gi(k) \]

For a \( 2N \)-point frequency domain sequences \( G(k) \) derived from a \( 2N \)-point real-valued sequences, perform the following steps for the IDFT of \( G(k) \).

Step 1: Initialize \( A^*(k) \)'s and \( B^*(k) \)'s.

As with the forward DFT, this step is usually performed only once during a power-up or initialization sequence. The values can be pre-stored in a boot ROM or computed. In either case, once the values are generated, this step is no longer needed when performing the DFT.

Because \( A^*(k) \) and \( B^*(k) \) are the complex conjugate of \( A(k) \) and \( B(k) \), respectively, each can be derived from the \( A(k) \)'s and \( B(k) \)'s. The following pseudo code is used to generate them.

for \( k = 0, 1, \ldots, N-1 \)

\[ A^*i(k) = \cos(\pi k/N) \]
\[ A^*r(k) = 1 - \sin(\pi k/N) \]
\[ B^*i(k) = -\cos(\pi k/N) \]
\[ B^*r(k) = 1 + \sin(\pi k/N) \]

Or, if \( A(k) \) and \( B(k) \) are already generated, you can use the following pseudo code.

for \( k = 0, 1, \ldots, N-1 \)

\[ A^*i(k) = -Ai(k) \]
\[ A^*r(k) = Ar(k) \]
\[ B^*i(k) = -Bi(k) \]
\[ B^*r(k) = Br(k) \]
Step 2: Let $G(k)$ be a $2N$-point complex valued sequence derived from a real-valued sequence $g(n)$.

We want to get back $g(n)$ from $G(k) \rightarrow g(n) = \text{IDFT}[G(k)]$. However, we want to apply the same techniques we applied with the forward DFT; that is, use an $N$-point IFFT. This can be accomplished using the following equations.

\[ G(N) = G(0) \]

for $k=0,1,\ldots,N-1$

\[
X_r(k) = G_r(k)A_r(k) - G_i(k)(-A_i(k)) + G_r(N-k)B_r(k) + G_i(N-k)(-B_i(k))
\]

\[
X_i(k) = G_i(k)A_r(k) + G_r(k)(-A_i(k)) + G_r(N-k)(-B_i(k)) - G_i(N-k)B_r(k)
\]

Step 3: Perform the $N$-point inverse DFT of $X(k)$.

\[ x(n) = x_1(n) + jx_2(n) = \text{IDFT}[X(k)] \]

NOTE:
The IDFT can be any method but must have an output in normal order.

Step 4: $g(n)$ can then be found from $x(n)$.

for $n = 0, 1, \ldots, N$

\[
g(2n) = x_1(n)
\]

\[
g(2n+1) = x_2(n)
\]

Appendix C contains C implementations of the outlined DFT and IDFT algorithms.
Before we discuss how to efficiently implement the real-valued FFT algorithms on the 'C62xx, it is helpful to take a brief look at the 'C62xx architecture and code development tools. The TMS320C62xx fixed-point processors are based on a 256-bit advanced VLIW CPU with eight functional units, including two multipliers and six ALUs. The CPU can execute up to eight 32-bit instructions per cycle. With an instruction clock frequency of 200 MHz and greater, 'C62xx peak performance starts at 1600 MIPS.

The 'C62xx processor consists of three main parts:

- CPU
- Peripherals
- Memory

Figure 1 shows a block diagram of the first device in this family, the TMS320C6201 DSP.

*Figure 1. TMS320C6201 DSP Block Diagram*
This discussion focuses on the CPU, or core, of the device. The ‘C62xx CPU is the central building block of all TMS320C62xx devices and features two data paths where processing occurs. Each data path has four functional units (.L, .S, .M, .D) along with a register file containing 16 32-bit general purpose registers.

The ‘C62xx is a load store architecture in which all functional units obtain operands from a register file rather than directly from memory.

The .D units load/store data from and to memory from the register file with an address reach of 32 bits. The ‘C62xx architecture is also byte addressable: the .D units can load or store data in either 8 bits (byte), 16 bits (halfword), or 32 bits (word). In addition, the .D units can perform 32-bit addition and subtraction and address calculations.

The .M units perform multiplication, featuring a 16-bit by 16-bit multiplier that produces a 32-bit result. Additional multiplier features include the ability to select either the 16 MSBs (most significant bits) or 16 LSBs (least significant bits) of a register operand and optionally left shift the multiplier output by one with saturation.

The .S units perform branches and shifting primarily but also perform bit field operations such as extract, set and clear bit fields, as well as 32-bit logical operations and 32-bit addition and subtraction. Another advanced feature of each .S unit is the ability to split its ALU to perform two 16-bits adds or subtracts in a single cycle.

Although the .S and .D units perform ALU functions, the .L unit is the main ALU for the CPU, performing both 32-bit and 40-bit integer arithmetic. The .L unit also features saturation logic, comparison instructions, and bit counting and can perform 32-bit logical operations. In support of the eight functional units, the CPU has a program fetch unit; instruction dispatch unit; instruction decode unit; control registers; control logic; and test, emulation and interrupt logic.

The ‘C62xx features a state of the art software development environment. A very efficient C compiler along with a linear assembly optimizer allows fast time to market through ease of use. Its orthogonal RISC-like CPU architecture makes the ‘C62xx CPU a very good C compiler target. Combined with TI's compiler expertise, these features make the ‘C62xx compiler the most efficient DSP compiler on the market today.
Because of its efficiency, most ‘C62xx coding can be done in C. However, as with many other DSPs, some tasks or routines require assembly coding to achieve the highest performance possible.

As a result, TI has developed a new tool called the assembly optimizer that makes assembly language coding easier and faster. The assembly optimizer allows you to write linear assembly code (no parallel instructions) without assigning registers to operands.

The assembly optimizer accepts this input syntax and generates an assembly language output that parallelizes the linear instructions and assigns registers to operands. This relieves the assembly language programmer of the following responsibilities:

- Determining which instructions can be executed in parallel
- Knowing how to position code to avoid delay slot conflicts
- Keeping track of which registers are live or free

The ‘C62xx assembler is also included in the code development tool set.

Figure 2 shows the process flow to develop code for the ‘C62xx.
Figure 2. Code Development Flow Chart
In phase 1 of the code development process, TI recommends that the algorithm first be implemented in C, which serves the following purposes:

- Provides an easy way to verify the functionality of an algorithm
- Provides a working model to verify results of optimized versions
- May meet your efficiency requirements and thus completes your implementation

If phase 1 fails to meet your performance requirements, you may need to proceed to phase 2 to refine and optimize your C code. The process includes modifying your C code for efficiency using the C code optimization methods. This section offers a brief overview of the C code optimization methods (for a more detailed explanation, see the TI TMS320C62xx Programmers Guide).

One of the easiest methods used to optimize your C code is the C compilers' optimizer, evoked using compiler options. Some of the most commonly used optimizer options are: -o3, -pm, -mt, and -x2 (see the TI TMS320C6x Optimizing C Compiler User’s Guide for a list of available compiler optimization options and usage).

Other methods to optimize C code for efficiency involve modifying your C code. One very effective method uses compiler intrinsics – special functions that map directly to inlined ‘C62xx instructions. Intrinsic functions typically allow you to use a ‘C62xx specific feature that is not directly expressible in C, such as .L unit saturation.

Other effective optimization techniques include

- Loop unrolling
- Software pipelining
- Trip count specification
- Using the const keyword to eliminate memory dependencies

All of these methods produce very efficient C code. Nevertheless, the compiler still may not produce the efficiency required. In this case, phase 3 may be required. Phase 3 uses the assembly optimizer and/or the assembler to generate ‘C62xx assembly code. By far the easiest and recommended route is the assembly optimizer (assembly optimizer usage is outlined in detail in the TI TMS320C6x Optimizing C Compiler User’s Guide. In addition, the TI TMS320C6200 Assembly Language Tools User’s Guide outlines assembler usage).
A recommended approach for using either assembly method is to implement assembly routines as C callable assembly functions.

**CAUTION:**

Use caution when implementing C callable assembly routines so that you do not disrupt the C environment and cause a program to fail. The TI *TMS320C6x Optimizing C Compiler User’s Guide* details the register, stack, calling and return requirements of the ‘C62xx runtime environment. TI recommends that you read the material covering these requirements before implementing a C callable assembly language function.
Implementation and Optimization of Real-Valued DFTs

Appendix C contains the source code listings for C implementation of the two efficient methods for performing the DFT of real-valued sequences outlined in this application report. Each implementation fits into phase 1 of the code development flowchart shown in Figure 2.

The primary purpose of this particular implementation is to verify the functionality of split operation algorithm implementations and provide a known good model to compare against optimized versions. Another benefit is that this implementation is generic C code and thus can be easily ported to other DSPs or CPUs featuring C compilers.

Because the primary focus of this application report is the split operations used in the efficient computation of DFTs, the C implementation is not efficient with respect to the other operations involved in the computation of real-valued DFTs. For example, the direct form of the DFT is implemented rather than a more computationally efficient FFT. Optimizing the C code to yield better performance is addressed in appendix D.

Example 1 and Example 2 show the compiler usage for building the executable files for these implementations.

Example 1. Efficient Computation of the DFT of a 2N-Point Real Sequence

```bash
cl6x -g vectors.asm realdft1.c split1.c data1.c dft.c -z -o test1.out -l rts6201.lib lnk.cmd
```

Example 2. Efficient Computation of the DFT of Two Real Sequences

```bash
cl6x -g vectors.asm realdft2.c split2.c data2.c dft.c -z -o test2.out -l rts6201.lib lnk.cmd
```

The example compiler usage results in two executable files that can be loaded into the ‘C62xx device simulator and run:

- test1.out
- test2.out

The –g option used in the above compiler usage tells the compiler to build the code with debug information. This means that the compiler does not use the optimizer but allows the code to be easily viewed by the debugger.
First time users of the ‘C62xx are encouraged to try different compiler options and compare the effects of each on code performance (for benchmarking code on the ‘C62xx debugger, see the TI TMS320C62x C Source Debugger User’s Guide).

Appendix D contains the source code listings for optimized C implementations of the two efficient methods for performing the DFT of real-valued sequences outlined in this application report. These implementations apply to phase 2 of the code development flowchart shown in Figure 2.

For this implementation, the C code is refined to yield better performance. Not all C optimization techniques outlined in this application report have been implemented. This is so that the C code remains generic and can be ported easily to other DSPs. However, you can easily apply other ‘C62xx C optimization techniques to increase performance. The following optimizations are implemented in Appendix D.

- The DFT is replaced with a radix-4 FFT, yielding a large computational savings as the number of data samples to be transformed increases. The radix-4 FFT restricts the size to a power of 4.
- Split operation tables and FFT twiddle factors are generated using pre-generated look-up tables instead of the run-time support functions sin() and cos(). This reduces the number of cycles required for the setup code.
- The code is organized as a series of functions to separate the independent tasks so they could be easily and independently optimized.

Example 3 and Example 4 show the compiler usage for building executable files for these implementations.

Example 3. Efficient Computation of the DFT of a 2N-Point Real Sequence

```
c16x -g vectors.asm data1.c digitgen.c digit.c radix4.c realdft3.c split1.c splitgen.c -z -o test3c.out -l rts6201.lib lnk.cmd
```

Example 4. Efficient Computation of the DFT of Two Real Sequences

```
c16x -g vectors.asm data2.c realdft4.c split2.c radix4.c digit.c digitgen.c -z -o test4c.out -l rts6201.lib lnk.cmd
```

The result of the above compiles is two executables:

- test3c.out
- test4c.out
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

The executables can be loaded into the ‘C62xx device simulator and run. The same restrictions that apply to the executables in Appendix C apply to these executables.

Appendix E contains ‘C62xx assembly language source code listings. These implementations fit into phase 3 of the code development flowchart shown in Figure 2. Each assembly listing contains a ‘C62xx C callable assembly language function that replaces an equivalent C function shown in Appendix D. The following list includes functions implemented in assembly.

- **split1.asm** The C callable assembly language function that implements the split routine for the efficient computation of the DFT of two real sequences algorithm.
- **split2.asm** The C callable assembly language function that implements the split routine for the efficient computation of the DFT of 2N-point real sequence.
- **radix4.asm** Replaces radix4.c. A C callable assembly language function that implements the radix-4 FFT.
- **digit.asm** Replaces digit.c. A C callable assembly language function that implements the digit reversal for the radix-4 FFT.

Because each of the above routines is functionally equivalent in C and assembly, no modification of other functions in Appendix D is required to use them. All that must be changed to use these functions is the way in which we build the executables. Example 5 and Example 6 show how to build the executables with the assembly versions.

**Example 5. Efficient Computation of the DFT of a 2N-Point Real Sequence**

```
cl6x -g vectors.asm data1.c digitgen.c digit.asm radix4.asm realdft3.c split1.asm splitgen.c -z -o test3a.out -l rts6201.lib lnk.cmd
```

**Example 6. Efficient Computation of the DFT of Two Real Sequences**

```
cl6x -g vectors.asm data2.c realdft4.c split2.asm radix4.asm digit.asm digitgen.c -z -o test4a.out -l rts6201.lib lnk.cmd
```

The result of the compiles shown in Example 5 and Example 6 is two executables:

- test3a.out
- test4a.out

These can be loaded into the ‘C62xx device simulator and run. The same restrictions that apply to the executables in Appendix C apply to these executables.
Summary

This application report examined the theory and implementation of two efficient methods for computing the DFT of real-valued sequences. The implementation was presented in both C and ‘C62xx assembly language. As this application report reveals, a large computational savings can be achieved using these methods on real-valued sequences rather using complex-valued DFTs or FFTs. Moreover, the TMS320C62xx CPU performs well when implementing these algorithms in either C or assembly.
Appendix A. Derivation of Equations Used to Compute the DFT/IDFT of Two Real Sequences

This appendix provides a detailed derivation of the equations used to compute the FFT/IDFT of two real sequences using one complex DFT/IDFT.

**Forward Transform**

Assume $x_1(n)$ and $x_2(n)$ are real valued sequences of length $N$, and let $x(n)$ be a complex-valued sequence defined as

$$x(n) = x_1(n) + jx_2(n) \quad 0 \leq n \leq N - 1 \quad (A1)$$

The DFT operation is linear, thus the DFT of $x(n)$ may be expressed as:

$$X(k) = X_1(k) + jX_2(k) \quad 0 \leq k \leq N - 1 \quad (A2)$$

We can express the sequences $x_1(n)$ and $x_2(n)$ in terms of $x(n)$ as follows:

$$x_1(n) = \frac{x(n) + x^*(n)}{2}$$

where $^*$ is the complex conjugate operator

$$x_2(n) = \frac{x(n) - x^*(n)}{2j}$$

The following shows that these equalities are true:

$$\frac{x(n) + x^*(n)}{2} = \frac{x_1(n) + jx_2(n) + x_1(n) - jx_2(n)}{2} = x_1(n) \quad (A4)$$

$$\frac{x(n) - x^*(n)}{2j} = \frac{x_1(n) + jx_2(n) - x_1(n) + jx_2(n)}{2j} = x_2(n)$$

Therefore we can express the DFT of $x_1(n)$ and $x_2(n)$ in terms of $x(n)$ as shown below:

$$X_1(k) = \text{DFT}[x_1(n)] = \frac{1}{2} \{\text{DFT}[x(n)] + \text{DFT}[x^*(n)]\} \quad (A5)$$

$$X_2(k) = \text{DFT}[x_2(n)] = \frac{1}{2j} \{\text{DFT}[x(n)] - \text{DFT}[x^*(n)]\}$$
From the complex property of the DFT we know the following is true:

If \( x(n) \xrightarrow{\text{DFT}_N} X(k) \), then \( x^*(n) \xrightarrow{\text{DFT}_N} X^*(N-k) \)

Thus, we can express \( X_1(k) \) and \( X_2(k) \) as follows:

\[
X_1(k) = \frac{1}{2} [x(k) + X^*(N-k)] \\
X_2(k) = \frac{1}{2j} [X(k) - X^*(N-k)]
\]

(A6)

From the above equations, we can see that by performing a single DFT on the complex-valued sequence \( x(n) \), we have obtained the DFT of two real-valued sequences with only a small amount of additional computation in calculating \( X_1(k) \) and \( X_2(k) \) from \( X(k) \).

In addition, because \( x_1(n) \) and \( x_2(n) \) are real-valued sequences, \( X_1(k) \) and \( X_2(k) \) has complex conjugate symmetry - \( X_1(N-k) = X_1^*(k) \) and \( X_2(N-k) = X_2^*(k) \), thus we only need to compute \( X_1(k) \) and \( X_2(k) \) for \( k = 0, 1, 2, \ldots, N/2 \).

\[
X_1(k) = \frac{1}{2} [X(k) + X^*(N-k)] \\
X_1(N-k) = X_1^*(k) \\
X_2(k) = \frac{1}{2j} [X(k) - X^*(N-k)] \\
X_2(N-k) = X_2^*(k)
\]

(A7)

To implement these equations, it is helpful if we express them in terms of their real and imaginary terms.

\[
X_1(k) = \frac{1}{2} \{Xr(k) + jXi(k) + Xr(N-k) - jXi(N-k)\} \\
= \frac{1}{2} \{(Xr(k) + Xr(N-k)) + j(Xi(k) - Xi(N-k))\}
\]

or,

\[
X_1r(k) = \frac{1}{2} \{Xr(k) + Xr(N-k)\} \\
X_1i(k) = \frac{1}{2} \{Xi(k) - Xi(N-k)\}
\]

(A8)
Similarly, it can be shown that

\[ X_r(k) = \frac{1}{2} \{X_i(k) + X_i(N-k)\} \]

\[ k = 0, 1, \ldots, N/2 \quad (A9) \]

with \( X(N) = X(0) \)

\[ X_i(k) = -\frac{1}{2} \{X_r(k) - X_r(N-k)\} \]

There are two special cases with the above equations, \( k = 0 \) and \( k = N/2 \). For \( k = 0 \):

\[ X_r(0) = \frac{1}{2} \{X_r(0) + X_r(N)\} \]

\[ X_i(0) = \frac{1}{2} \{X_i(0) - X_i(N)\} \]

\[ X_r(0) = \frac{1}{2} \{X_i(0) + X_i(N)\} \]

\[ X_i(0) = -\frac{1}{2} \{X_r(0) - X_r(N)\} \]

(A10)

Because of the periodicity property of the DFT, we know \( X(k + N) = X(k) \). Therefore \( X_r(0) = X_r(N) \) and \( X_i(0) = X_i(N) \). Using this property, the above equations can be expressed as follows:

\[ X_r(0) = X_r(0) \]

\[ X_i(0) = 0 \]

\[ X_r(0) = X_i(0) \]

\[ X_i(0) = X_r(0) \]

(A11)

For \( k = N/2 \):

\[ X_r(N/2) = \frac{1}{2} \{X_r(N/2) + X_r(N/2)\} \]

\[ X_i(N/2) = \frac{1}{2} \{X_i(N/2) - X_i(N/2)\} \]

\[ X_r(N/2) = \frac{1}{2} \{X_i(N/2) + X_i(N/2)\} \]

\[ X_i(N/2) = -\frac{1}{2} \{X_r(N/2) - X_r(N/2)\} \]

or

\[ (A12) \]
Thus, (A8) and (A9) must be computed only for \( k = 1, 2, \ldots, N/2 - 1 \).

### Inverse Transform

We can use a similar method to obtain the IDFT. We know \( X_1(k) \) and \( X_2(k) \). We want to express \( X(k) \) in terms of \( X_1(k) \) and \( X_2(k) \). Recall, the relationship between \( x_1(n) \), \( x_2(n) \) and \( x(n) \) is

\[
x(n) = x_1(n) + jx_2(n).
\]

Since the DFT operator is linear

\[
X(k) = X_1(k) + jX_2(k).
\]

Thus, \( X(k) \) can be found by the following equations:

\[
X_r(k) = X_1r(k) - X_2i(k)
\]

\[
X_i(k) = X_1i(k) + X_2r(k)
\]

\( x(n) \) can then be found by taking the inverse transform of \( X(k) \).

\[
x(n) = \text{IDFT}[X(k)]
\]

From \( x(n) \), we can get \( x_1(n) \) and \( x_2(n) \).

\[
x_1(n) = x_r(n)
\]

\[
x_2(n) = x_i(n)
\]
Appendix B. Derivation of Equations Used to Compute the DFT/IDFT of a Real Sequence

This appendix details the derivation of the equations used to compute the DFT/IDFT of a 2N-length real-valued sequence using an N-length complex DFT/IDFT.

Forward Transform

Assume $g(n)$ is a real-valued sequence of 2N points. The following shows how to obtain the 2N-point DFT of $g(n)$ using an N-point complex DFT.

Let

$$
x_1(n) = g(2n)
$$

$$
x_2(n) = g(2n + 1)
$$

We have subdivided a 2N-point real sequence into two N-point sequences. We now can apply the same method shown in Appendix A.

Let $x(n)$ be the N-point complex-valued sequence.

$$
x(n) = x_1(n) + jx_2(n)
$$

0 $\leq n \leq N - 1$  \hspace{1cm} (B2)

From the results shown in Appendix A, we have

$$
X_1(k) = \frac{1}{2} \left\{ X(k) + X^*(N - k) \right\}
$$

$k = 0, 1, ..., N - 1$  \hspace{1cm} (B3)

$$
X_2(k) = \frac{1}{2j} \left\{ X(k) - X^*(N - k) \right\}
$$

We now express the 2N-point DFT in terms of two N-point DFTs.

$$
G(k) = \text{DFT}[g(n)] = \text{DFT}[g(2n) + g(2n + 1)] = \text{DFT}[g(2n)] + \text{DFT}[g(2n + 1)]
$$

$$
= \sum_{n=0}^{N-1} g(2n)W_N^{2nk} + \sum_{n=0}^{N-1} g(2n + 1)W_N^{(2n+1)k}
$$

$k = 0, 1, ..., N - 1$  \hspace{1cm} (B4)

$$
= \sum_{n=0}^{N-1} x_1(n)W_N^{nk} + W_N^{k} \sum_{n=0}^{N-1} x_2(n)W_N^{nk}
$$

Thus,
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

\[ G(k) = X_1(k) + W_{2N}^k X_2(k) \quad k = 0,1,...,N-1 \quad (B5) \]

Using equation (B3), we can express \( G(k) \) in terms of \( X(k) \).

\[
G(k) = \frac{1}{2} \{ X(k) + X^*(N-k) \} + W_{2N}^k \frac{1}{2j} \{ X(k) - X^*(N-k) \}
\]

\[
= X(k)[\frac{1}{2} (1 - jW_{2N}^k)] + X^*(N-k)[\frac{1}{2} (1 + jW_{2N}^k)]
\]

Let

\[
A(k) = [\frac{1}{2} (1 - jW_{2N}^k)] \quad k = 0,1,...,N-1 \quad (B7)
\]

\[
B(k) = [\frac{1}{2} (1 + jW_{2N}^k)]
\]

Thus, \( G(k) \) can be expressed as follows:

\[
G(k) = X(k)A(k) + X^*(N-k)B(k) \quad k = 0,1,...,N-1 \quad (B8)
\]

Because \( x(n) \) is a real-valued sequence, we know that the DFT transform results will have complex conjugate symmetry. Also, because of the periodicity property of the DFT, we know \( X(k+N) = X(k) \); therefore, \( X(N) = X(0) \). Using these properties, we can find the other half of the DFT result.

Thus, we have computed the DFT of a 2N-point real sequence using one N-point complex DFT and additional computations.

To implement these equations, it is helpful to express them in terms of their real and imaginary terms.

\[
G(k) = (Xr(k) + j Xi(k))(Ar(k) + j Ai(k)) + (Xr(N-k) - j Xi(N-k))(Br(k) + j Bi(k))
\]

\[
k = 0, 1, ..., N-1 \quad (B9)
\]
Carrying out the multiplication, separating the real and imaginary terms, and applying the periodicity and complex conjugate properties, we have the following:

\[ Gr(k) = Xr(k)Ar(k) - Xi(k)Ai(k) + Xr(N-k)Br(k) + Xi(N-k)Bi(k) \]

\[ k = 0, 1, \ldots, N-1 \]

with \( X(N) = X(0) \)

\[ Gi(k) = Xi(k)Ar(k) + Xr(k)Ai(k) + Xr(N-k)Bi(k) - Xi(N-k)Br(k) \]

(B10)

\[ Gr(k) = Xr(0) - Xi(0) \]

\[ k = N \]

\[ Gi(k) = 0 \]

\[ Gr(2N-k) = Gr(k) \]

\[ k = 1, 2, \ldots, N-1 \]

\[ Gi(2N-k) = -Gi(k) \]

**Inverse Transform**

We will now derive the equations for the IDFT of a \( 2N \)-point complex sequence derived from a real sequence using an \( N \)-point complex IDFT. We express the \( N \)-point complex sequence \( X(k) \) in terms of the \( 2N \)-point complex sequence \( G(k) \). Once \( X(k) \) is known, \( x(n) \) can be found by taking the IDFT of \( X(k) \). Once \( x(n) \) is known, \( g(n) \) follows.

Equation (B8) can be re-written as follows:

\[ G(k) = X(k)A(k) + X*(N-k)B(k) \]

\[ k = 0, 1, \ldots, N/2 \]

\[ G(N-k) = X(N-k)A(N-k) + X*(k)B(N-k) \]

\[ k = 0, 1, \ldots, N/2 - 1 \]

(B11)

where

\[ A(N-k) = \left[ \frac{1}{2} (1 - jW_{2N}^{-N+k}) \right] = \left[ \frac{1}{2} (1 + jW_{2N}^{-k}) \right] = A^*(k) \]

(B12)

\[ B(N-k) = \left[ \frac{1}{2} (1 + jW_{2N}^{-N+k}) \right] = \left[ \frac{1}{2} (1 - jW_{2N}^{-k}) \right] = B^*(k) \]

The above equalities can be shown to be true by recalling the following definition and substituting appropriately for \( k \).

\[ W_{2N}^{k} = e^{-j2\pi k/2N} = \cos(2\pi k/2N) - j \sin(2\pi k/2N) \]

(B13)
We would like to make the ranges of $k$ for $G(k)$ and $G(N-k)$ the same. Look at $G(N/2)$:

$$G(N/2) = X(N/2)A(N/2) + X^*(N/2)B(N/2) \quad \text{(B14)}$$

$$= X(N/2)[\frac{1}{2}(1 - jW_{2N}^{N/2})] + X^*(N/2)[\frac{1}{2}(1 + jW_{2N}^{N/2})]$$

But from (B13), we see that

$$W_{2N}^{N/2} = -j \quad \text{(B15)}$$

Therefore:

$$G(N/2) = X^*(N/2) \quad \text{(B16)}$$

Now we can express $G(k)$ and $G(N-k)$ with the same ranges of $k$, and along with (B12) and (B14) we have:

$$G(k) = X(k)A(k) + X^*(N-k)B(k) \quad k = 0,1,\ldots,N/2-1 \quad \text{(B17)}$$

$$G(N-k) = X(N-k)A^*(k) + X^*(k)B^*(k) \quad k = 0,1,\ldots,N/2-1 \quad \text{(B18)}$$

$$G(N/2) = X^*(N/2)$$

From (B17) and (B18) you can see we have two equations and two unknowns, $X(k)$ and $X(N-k)$. We can use some algebra tricks to come up with equations for $X(k)$ and $X(N-k)$. If we multiply both sides of (B17) with $A(k)$, complex conjugate both sides of (B18), then multiply both sides by $B(k)$, we have the following:

$$G(k)A(k) = X(k)A(k)A(k) + X^*(N-k)B(k)A(k) \quad \text{(B19)}$$

$$-G^*(N-k)B(k) = X(k)B(k)B(k) + X^*(N-k)B(k)A(k)$$

$$G(k)A(k) - G^*(N-k)B(k) = X(k)\{A(k)A(k) - B(k)B(k)\} \quad k = 0,1,\ldots,N/2-1$$

Solving for $X(k)$

$$X(k) = \frac{G(k)A(k) - G^*(N-k)B(k)}{A(k)A(k) - B(k)B(k)} \quad k = 0,1,\ldots,N/2-1 \quad \text{(B20)}$$
Equation (B20) can be simplified as follows:

\[ A(k)A(k) = \left[ \frac{1}{2}(1 - jW_{2N}^k) \right] \left[ \frac{1}{2}(1 - jW_{2N}^k) \right] = \frac{1}{4} \left[ 1 - 2jW_{2N}^k - W_{2N}^k \right] \quad (B21) \]

\[ B(k)B(k) = \left[ \frac{1}{2}(1 + jW_{2N}^k) \right] \left[ \frac{1}{2}(1 + jW_{2N}^k) \right] = \frac{1}{4} \left[ 1 + 2jW_{2N}^k - W_{2N}^k \right] \quad (B22) \]

\[ A(k)A(k) - B(k)B(k) = -jW_{2N}^k \]

\[ X(k) = \frac{G(k)A(k) - G^*(N-k)B(k)}{-jW_{2N}^k} \quad k = 0, 1, \ldots, N/2 - 1 \quad (B23) \]

Similarly, if we multiply both sides of (B18) with \( A^*(k) \), conjugate both sides of (B17), then multiply both sides by \( B^*(k) \), we have the following:

\[ G^*(k)B^*(k) = X^*(k)A^*(k)B^*(k) + X(N-k)B^*(k)B^*(k) \]

\[ G(N-k)A^*(k) = X^*(k)A^*(k)B^*(k) + X(N-k)A^*(k)A^*(k) \]

\[ G^*(k)B^*(k) - G(N-k)A^*(k) = X(N-k)[A^*(k)A^*(k) - B^*(k)B^*(k)] \quad (B24) \]

\[ k = 0, 1, \ldots, N/2 - 1 \]

Solving for \( X(N-k) \):

\[ X(N-k) = \frac{G^*(k)B^*(k) - G(N-k)A^*(k)}{A^*(k)A^*(k) - B^*(k)B^*(k)} \quad k = 0, 1, \ldots, N/2 - 1 \quad (B25) \]

Equation (B25) can be simplified as follows:

\[ A^*(k)A^*(k) = \left[ \frac{1}{2}(1 + jW_{2N}^{-k}) \right] \left[ \frac{1}{2}(1 + jW_{2N}^{-k}) \right] = \frac{1}{4} \left[ 1 + 2jW_{2N}^{-k} - W_{2N}^{-k} \right] \quad (B26) \]

\[ B^*(k)B^*(k) = \left[ \frac{1}{2}(1 - jW_{2N}^{-k}) \right] \left[ \frac{1}{2}(1 - jW_{2N}^{-k}) \right] = \frac{1}{4} \left[ 1 - 2jW_{2N}^{-k} - W_{2N}^{-k} \right] \quad (B27) \]

\[ A^*(k)A^*(k) - B^*(k)B^*(k) = jW_{2N}^{-k} \quad (B28) \]

\[ X(N-k) = \frac{G^*(k)B^*(k) - G(N-k)A^*(k)}{jW_{2N}^{-k}} \quad k = 0, 1, \ldots, N/2 - 1 \quad (B29) \]
It can be shown that

\[
\frac{A(k)}{-jW_{2N}^k} = A^*(k) \quad \text{and} \quad \frac{B(k)}{-jW_{2N}^k} = B^*(k)
\] (B30)

\[
\frac{A^*(k)}{jW_{2N}^k} = A(k) \quad \text{and} \quad \frac{B^*(k)}{jW_{2N}^k} = B(k)
\] (B31)

Making these substitutions, we get:

\[
X(k) = G(k)A^*(k) + G^*(N - k)B^*(k)
\]

\[
k = 0,1,\ldots,N/2 - 1
\]

\[
X(N - k) = G^*(k)B(k) + G(N - k)A(k)
\] (B32)

\[
X(N/2) = G^*(N/2)
\]

For equation (B32), if we make the following substitutions, along with replacing \(k\) with \(N/2 - k\),

\[
A(N - k) = A^*(k) \quad \text{and} \quad B(N - k) = B^*(k)
\] (B33)

we can see that \(X(k)\) can be expressed a single equation.

\[
X(k) = G(k)A^*(k) + G^*(N - k)B^*(k)
\]

\[
k = 0,1,\ldots,N - 1
\]

\[
G(N) = G(0)
\] (B34)

Now in terms of implementing these equations, it is helpful to express them in terms of their real and imaginary terms.

\[
X(k) = (Gr(k) + jGi(k))(Ar(k) - jAi(k)) + (Gr(N - k) - jGi(N - k))(Br(k) - jBi(k))
\]

\[
k = 0,1,\ldots,N - 1
\]

with \(G(N) = G(0)\)

(B35)

Carrying out the multiplication and separating the real and imaginary terms, we have the following:

\[
Xr(k) = Gr(k)Ar(k) + Gi(k)Ai(k) + Gr(N - k)Br(k) - Gi(N - k)Bi(k)
\]

\[
k = 0,1,\ldots,N - 1
\]

with \(G(N) = G(0)\)

\[
Xi(k) = Gi(k)Ar(k) - Gr(k)Ai(k) - Gr(N - k)Bi(k) - Gi(N - k)Br(k)
\] (B36)
Now we have formed the complex sequence with which we can use an $N$-point complex DFT to obtain $x(n)$, which we then can use to get $g(n)$.

$$x(n) = xr(n) + jxi(n) = \text{IDFT}[X(k)]$$

$$g(2n) = xr(n) \quad \quad \quad \quad \quad n = 0, 1, \ldots, N - 1$$

$$g(2n + 1) = xi(n)$$
Appendix C. C Implementations of the DFT of Real Sequences

This appendix contains C implementations of the efficient methods for performing the DFT of real-valued sequences.

Implementation Notes

The following lists usage, assumptions, and limitations of the code.

**Data format**

All data and state variables are 16-bit signed integers (shorts). In this example, the decimal point is assumed to be between bits 15 and 14, thus the Q15 data format. For complex data and variables, the real and imaginary components are both Q15 numbers. From this data format, you can see that this code was developed for a fixed-point processor.

**Memory**

Complex data is stored in memory in imaginary/real pairs. The imaginary component is stored in the most significant halfword (16 bits) and the real component is stored in the least significant halfword, unless otherwise noted.

**Endianess**

The code is presented and tested in little endian format. Some modification to the code is necessary for big endian format.

**Overflow**

No overflow protection or detection is performed.

**File**

**Description**

- **realdft1.c**
  DFT of a $2N$-point real sequence main program
- **split1.c**
  Split function for the DFT of a $2N$-point real sequence
- **data1.c**
  Sample data
- **params1.h**
  Header file, for example
- **realdft2.c**
  DFT of a two $N$-point real sequence main program
- **split2.c**
  Split function for the DFT of two $N$-point real sequence
<table>
<thead>
<tr>
<th>File</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>data2.c</td>
<td>Sample data</td>
</tr>
<tr>
<td>params2.h</td>
<td>Header file, for example</td>
</tr>
<tr>
<td>dft.c</td>
<td>Direct implementation of the DFT function</td>
</tr>
<tr>
<td>params.h</td>
<td>Header file</td>
</tr>
<tr>
<td>vectors.asm</td>
<td>Reset vector assembly source.</td>
</tr>
<tr>
<td>lnk.cmd</td>
<td>Example linker command file.</td>
</tr>
</tbody>
</table>
Example 7. realdft1.c File

FILENAME
realdft1.c - C source for an example implementation of the DFT/IDFT of a 2N-point real sequences using one N-point complex DFT/IDFT.

DESCRIPTION
This program is an example implementation of an efficient way of computing the DFT/IDFT of a real-valued sequence.

In many applications, the input is a sequence of real numbers. If this condition is taken into consideration, additional computational savings can be achieved because the FFT of a real sequence has some symmetrical properties. The DFT of a 2N-point real sequence can be efficiently computed using a N-point complex DFT and some additional computations.

The following steps are required in the computation of the FFT of a real valued sequence using the split function:

1. Let g(n) be a 2N-point real sequence. From g(n), form the N-point complex valued sequence, x(n) = x1(n) + jx2(n), where x1(n) = g(2n) and x2(n) = g(2n + 1).

2. Perform an N-point complex FFT on the complex valued sequence x(n) -> X(k) = DFT{x(n)}. Note the FFT can be any DFT method, such as radix-2, radix-4, mixed radix, direct implementation of the DFT, etc. However, the DFT output must be in normal order.

3. The following additional computations are used to get G(k) from X(k)

\[ Gr(k) = X_r(k)A_r(k) - X_i(k)A_i(k) + X_r(N-k)B_r(k) + X_i(N-k)B_i(k) \]
\[ Gi(k) = X_i(k)A_r(k) + X_r(k)A_i(k) + X_r(N-k)B_i(k) - X_i(N-k)B_r(k) \]

Note, only N-points of the 2N-point sequence of G(k) is computed in the above equations. Because the DFT of a real-sequence has symmetric properties, we can easily compute the remaining N points of G(k) with the following equations.

\[ Gr(N) = X_r(0) - X_i(0) \]
\[ Gi(N) = 0 \]
\[ Gr(2N-k) = Gr(k) \]
\[ Gi(2N-k) = -Gi(k) \]

As you can see, the above equations assume that A(k) and B(k), which are sine and cosine coefficients, are pre-computed. The C-code can be used to initialize A(k) and B(k).

for(k=0; k<N; k++)
{
    A[k].imag = (short)(16383.0*(-cos(2*PI/(double)(2*N)*(double)k)));
    A[k].real = (short)(16383.0*(1.0 - sin(2*PI/(double)(2*N)*(double)k)));
    B[k].imag = (short)(16383.0*(cos(2*PI/(double)(2*N)*(double)k)));
    B[k].real = (short)(16383.0*(1.0 + sin(2*PI/(double)(2*N)*(double)k)));
}

The following steps are required in the computation of the IFFT of a complex valued frequency domain sequence, that was derived from a real sequence:
1. Let $G(k)$ be a $2N$-point complex valued sequence derived from a real valued sequence $g(n)$. We want to get back $g(n)$ from $G(k) \rightarrow g(n) = \text{IDFT}(G(k))$. However, we want to apply the same techniques as we did with the forward FFT, use a $N$-point IFFT. This can be accomplished by the following equations.

$$X_r(k) = G_r(k)I_a(k) - G_i(k)I_b(k) + G_r(N-k)I_b(k) - G_i(N-k)I_a(k) \quad k = 0, 1, \ldots, N-1$$

and $G(N) = G(0)$

$$X_i(k) = G_i(k)I_a(k) + G_r(k)I_b(k) + G_r(N-k)I_a(k) - G_i(N-k)I_b(k)$$

2. Perform the $N$-point inverse DFT of $X(k) \rightarrow x(n) = x_1(n) + jx_2(n) = \text{IDFT}(X(k))$. Note, the IDFT can be any method, but must have an output that is in normal order.

3. $g(n)$ can then be found from $x(n)$.

$$g(2n) = x_1(n) \quad n = 0, 1, \ldots, N-1$$

$$g(2n+1) = x_2(n)$$

As you can see, the above equations can be used for both the forward and inverse FFTs, however, the pre-computed coefficients are slightly different. The following C-code can be used to initialize $IA(k)$ and $IB(k)$.

```c
for(k=0; k<N; k++)
{
    IA[k].imag = -(short)(16383.0*(-cos(2*PI/(double)(2*N)*(double)k)));
    IA[k].real = (short)(16383.0*(1.0 - sin(2*PI/(double)(2*N)*(double)k)));
    IB[k].imag = -(short)(16383.0*(cos(2*PI/(double)(2*N)*(double)k)));
    IB[k].real = (short)(16383.0*(1.0 + sin(2*PI/(double)(2*N)*(double)k)));
}
```

Note, $IA(k)$ is the complex conjugate of $A(k)$ and $IB(k)$ is the complex conjugate of $B(k)$.

```c
#include <math.h>
#include "params1.h"
#include "params.h"

extern short g[];

void dft(int, COMPLEX *);
void split(int, COMPLEX *, COMPLEX *, COMPLEX *, COMPLEX *);

main()
{
    int n, k;

    COMPLEX x[NUMPOINTS+1]; /* array of complex DFT data */
    COMPLEX A[NUMPOINTS]; /* array of complex A coefficients */
    COMPLEX B[NUMPOINTS]; /* array of complex B coefficients */
    COMPLEX IA[NUMPOINTS]; /* array of complex A* coefficients */
    COMPLEX IB[NUMPOINTS]; /* array of complex B* coefficients */
    COMPLEX G[2*NUMPOINTS]; /* array of complex DFT result */

    /* Initialize A, B, IA, and IB arrays */

    for(k=0; k<NUMPOINTS; k++)
    {
        A[k].imag = (short)(16383.0*(-cos(2*PI/(double)(2*NUMPOINTS)*(double)k)));
        A[k].real = (short)(16383.0*(1.0 - sin(2*PI/(double)(2*NUMPOINTS)*(double)k)));
        B[k].imag = (short)(16383.0*(cos(2*PI/(double)(2*NUMPOINTS)*(double)k)));
        B[k].real = (short)(16383.0*(1.0 + sin(2*PI/(double)(2*NUMPOINTS)*(double)k)));
    }
```
B[k].real = (short)(16383.0*(1.0 + sin(2*PI/(double)(2*NUMPOINTS)*(double)k)));
IA[k].imag = -A[k].imag;
IA[k].real = A[k].real;
IB[k].imag = -B[k].imag;
IB[k].real = B[k].real;
}

/* Forward DFT */
/* From the 2N point real sequence, g(n), for the N-point complex sequence, x(n) */
for (n=0; n<NUMPOINTS; n++)
{
    x[n].imag = g[2*n + 1]; /* x2(n) = g(2n + 1) */
    x[n].real = g[2*n]; /* x1(n) = g(2n) */
}
/* Compute the DFT of x(n) to get X(k) -> X(k) = DFT{x(n)} */
dft(NUMPOINTS, x);
/* Because of the periodicity property of the DFT, we know that X(N+k)=X(k). */
x[NUMPOINTS].real = x[0].real;
x[NUMPOINTS].imag = x[0].imag;
/* The split function performs the additional computations required to get G(k) from X(k). */
split(NUMPOINTS, x, A, B, G);
/* Use complex conjugate symmetry properties to get the rest of G(k) */
G[NUMPOINTS].real = x[0].real - x[0].imag;
G[NUMPOINTS].imag = 0;
for (k=1; k<NUMPOINTS; k++)
{
    G[2*NUMPOINTS-k].real = G[k].real;
    G[2*NUMPOINTS-k].imag = -G[k].imag;
}
/* Inverse DFT - We now want to get back g(n) */
/* The split function performs the additional computations required to get X(k) from G(k). */
split(NUMPOINTS, G, IA, IB, x);
/* Take the inverse DFT of X(k) to get x(n). Note the inverse DFT could be any
IDFT implementation, such as an IFFT. */
/* The inverse DFT can be calculated by using the forward DFT algorithm directly
by complex conjugation - x(n) = (1/N)(DFT[X*(k)])*, where * is the complex conjugate
operator. */
/* Compute the complex conjugate of X(k). */
for (k=0; k<NUMPOINTS; k++)
{
    x[k].imag = -x[k].imag; /* complex conjugate X(k) */
}
/* Compute the DFT of X*(k). */
dft(NUMPOINTS, x);
/* Complex conjugate the output of the DFT and divide by N to get x(n). */

for (n=0; n<NUMPOINTS; n++)
{
    x[n].real = x[n].real/16;
    x[n].imag = (-x[n].imag)/16;
}

/* g(2n) = xr(n) and g(2n + 1) = xi(n) */

for (n=0; n<NUMPOINTS; n++)
{
    g[2*n] = x[n].real;
    g[2*n + 1] = x[n].imag;
}

return(0);
Example 8. split1.c File

```c
#include "params1.h"
#include "params.h"

void split(int N, COMPLEX *X, COMPLEX *A, COMPLEX *B, COMPLEX *G)
{
    int k;
    int Tr, Ti;
    for (k=0; k<N; k++)
    {
        Tr = (int)X[k].real * (int)A[k].real - (int)X[k].imag * (int)A[k].imag +
            (int)X[N-k].real * (int)B[k].real + (int)X[N-k].imag * (int)B[k].imag;
        G[k].real = (short)(Tr>>15);
        Ti = (int)X[k].imag * (int)A[k].real + (int)X[k].real * (int)A[k].imag +
            (int)X[N-k].real * (int)B[k].imag - (int)X[N-k].imag * (int)B[k].real;
        G[k].imag = (short)(Ti>>15);
    }
}
```

**FILE**

split1.c - This is the C source code for the implementation of the split routine which is the additional computations in computing the DFT of an 2N-point real-valued sequences using a N-point complex DFT.

**DESCRIPTION**

Computation of the DFT of 2N-point real-valued sequences can be efficiently computed using one N-point complex DFT and some additional computations. This function implements these additional computations which are shown below.

\[
Gr(k) = Xr(k)Ar(k) - Xi(k)Ai(k) + Xr(N-k)Br(k) + Xi(N-k)Bi(k) \\
k = 0, 1, \ldots, N-1 \\
\text{and } X(N) = X(0) \\
Gi(k) = Xi(k)Ar(k) + Xr(k)Ai(k) + Xr(N-k)Bi(k) - Xi(N-k)Br(k)
\]

Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

47
Example 9.  data1.c File

 cdecl
 FILE
data1.c - Sample data used in realdft1.c

 /******************************************************************************
 | array of real-valued input sequence, g(n) |
 ******************************************************************************/
 short g[] = {255, -35, 255, -35, 255, 255, 255, 255,
              255, 255, 255, 20, 255, 255, 255, 255,
              0, 0, 0, 0, 0, 0, 0, 0,
              0, 0, 0, 0, 0, 0, 0, 0};
Example 10. params1.h File

/* ******************************************
FILE
params1.h - This is the C header file for example real FFT
implementations.
*******************************************/

#define NUMDATA 32        /* number of real data samples */
#define NUMPOINTS NUMDATA/2 /* number of point in the DFT */
Example 11. realdft2.c File

/**
 * realdft2.c - C source for an example implementation of the DFT/IDFT of two N-point real sequences using one N-point complex DFT/IDFT.
 */

FILE
реалдft2.c - C source for an example implementation of the DFT/IDFT of two N-point real sequences using one N-point complex DFT/IDFT.

DESCRIPTION
This program is an example implementation of an efficient way of computing the DFT/IDFT of two real-valued sequences.

Assume we have two real-valued sequences of length N - x1[n] and x2[n]. The DFT of x1[n] and x2[n] can be computed with one complex-valued DFT of length N, as shown above, by following this algorithm.

1. Form the complex-valued sequence x[n] from x1[n] and x2[n]
   \[ x_r[n] = x_1[n] \quad \text{and} \quad x_i[n] = x_2[n], \quad 0, 1, \ldots, N-1 \]
   Note, if the sequences x1[n] and x2[n] are coming from another algorithm or a data acquisition driver, this step may be eliminated if these put the data in the complex-valued format correctly.

2. Compute \( X[k] = \text{DFT}(x[n]) \)
   This can be the direct form DFT algorithm, or an FFT algorithm. If using an FFT algorithm, make sure the output is in normal order - bit reversal is performed.

3. Compute the following equations to get the DFTs of x1[n] and x2[n].
   \[
   \begin{align*}
   X_1r[0] &= X_r[0] \\
   X_1i[0] &= 0 \\
   X_2r[0] &= X_i[0] \\
   X_2i[0] &= 0 \\
   X_1r[N/2] &= X_r[N/2] \\
   X_1i[N/2] &= 0 \\
   X_2r[N/2] &= X_i[N/2] \\
   X_2i[N/2] &= 0 \\
   \end{align*}
   \]
   for \( k = 1, 2, 3, \ldots, N/2-1 \)
   \[
   \begin{align*}
   X_1r[k] &= (X_r[k] + X_r[N-k])/2 \\
   X_1i[k] &= (X_i[k] - X_i[N-k])/2 \\
   X_1r[N-k] &= X_1r[k] \\
   X_1i[N-k] &= X_1i[k] \\
   X_2r[k] &= (X_i[k] + X_i[N-k])/2 \\
   X_2i[k] &= (X_r[N-k] - X_r[k])/2 \\
   X_2r[N-k] &= X_2r[k] \\
   X_2i[N-k] &= X_2i[k] \\
   \end{align*}
   \]

4. Form \( X[k] \) from \( X_1[k] \) and \( X_2[k] \)
   for \( k = 0, 1, \ldots, N-1 \)
   \[
   \begin{align*}
   x_r[k] &= X_1r[k] - X_2i[k] \\
   x_i[k] &= X_1i[k] + X_2r[k] \\
   \end{align*}
   \]

5. Compute \( x[n] = \text{IDFT}(X[k]) \)
   This can be the direct form IDFT algorithm, or an IFFT algorithm. If using
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

...an IFFT algorithm, make sure the output is in normal order - bit reversal is performed.

#include <math.h>    /* include the C RTS math library */
#include "params2.h" /* include file with parameters */
#include "params.h"  /* include file with parameters */

extern short x1[];
extern short x2[];

void dft(int, COMPLEX *);
extern void split2(int, COMPLEX *, COMPLEX *, COMPLEX *);

main()
{
    int n, k;
    COMPLEX X1[NUMDATA];     /* array of real-valued DFT output sequence, X1(k) */
    COMPLEX X2[NUMDATA];     /* array of real-valued DFT output sequence, X2(k) */
    COMPLEX x[NUMPOINTS+1];  /* array of complex DFT data, X(k) */

    /* Forward DFT */
    /* From the two N-point real sequences, x1(n) and x2(n), form the N-point complex sequence, x(n) = x1(n) + jx2(n) */
    for (n=0; n<NUMDATA; n++)
    {
        x[n].real = x1[n];
        x[n].imag = x2[n];
    }

    /* Compute the DFT of x(n), X(k) = DFT{x(n)}. Note, the DFT can be any DFT implementation such as FFTs. */
    dft(NUMPOINTS, x);

    /* Because of the periodicity property of the DFT, we know that X(N+k)=X(k). */
    x[NUMPOINTS].real = x[0].real;
    x[NUMPOINTS].imag = x[0].imag;

    /* The split function performs the additional computations required to get X1(k) and X2(k) from X(k). */
    split2(NUMPOINTS, x, X1, X2);

    /* Inverse DFT - We now want to get back x1(n) and x2(n) from X1(k) and X2(k) using one complex DFT */

    /* Recall that x(n) = x1(n) + jx2(n). Since the DFT operator is linear, X(k) = X1(k) + jX2(k). Thus we can express X(k) in terms of X1(k) and X2(k). */
    for (k=0; k<NUMPOINTS; k++)
    {
        
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

51
x[k].real = X1[k].real - X2[k].imag;
x[k].imag = X1[k].imag + X2[k].real;
}

/* Take the inverse DFT of X(k) to get x(n). Note the inverse DFT could be any
IDFT implementation, such as an IFFT. */
/* The inverse DFT can be calculated by using the forward DFT algorithm directly
by complex conjugation - x(n) = (1/N) DFT\{X*(k)}\}, where * is the complex conjugate
operator. */
/* Compute the complex conjugate of X(k). */
for (k=0; k<NUMPOINTS; k++)
{
    x[k].imag = -x[k].imag;
}
/* Compute the DFT of X*(k). */
dft(NUMPOINTS, x);
/* Complex conjugate the output of the DFT and divide by N to get x(n). */
for (n=0; n<NUMPOINTS; n++)
{
    x[n].real = x[n].real/16;
    x[n].imag = (-x[n].imag)/16;
}
/* x1(n) is the real part of x(n), and x2(n) is the imaginary part of x(n). */
for (n=0; n<NUMDATA; n++)
{
    x1[n] = x[n].real;
    x2[n] = x[n].imag;
}
return(0);
Example 12. split2.c File

/*****************************************************************************
FILE
split2.c - This is the C source code for the implementation of the
split routine which is the additional computations in computing the
DFT of two N-point real-valued sequences using one N-point complex DFT.
*****************************************************************************/

DESCRIPTION
Computation of the DFT of two N-point real-valued sequences can be
efficiently computed using one N-point complex DFT and some additional
computations. This function implements these additional computations
which are shown below.

\[ \begin{align*}
X1r[0] &= Xr[0] \\
X1i[0] &= 0 \\
X2r[0] &= Xi[0] \\
X2i[0] &= 0 \\
X1r[N/2] &= Xr[N/2] \\
X1i[N/2] &= 0 \\
X2r[N/2] &= Xi[N/2] \\
X2i[N/2] &= 0 \\
\end{align*} \]

\[ \text{for } k = 1, 2, 3, \ldots, N/2-1 \]
\[ \begin{align*}
X1r[k] &= (Xr[k] + Xr[N-k])/2 \\
X1i[k] &= (Xi[k] - Xi[N-k])/2 \\
X1r[N-k] &= X1r[k] \\
X1i[N-k] &= X1i[k] \\
X2r[k] &= (Xi[k] + Xi[N-k])/2 \\
X2i[k] &= (Xr[N-k] - Xr[k])/2 \\
X2r[N-k] &= X2r[k] \\
X2i[N-k] &= X2i[k] \\
\end{align*} \]

**************************************************************************/

#include "params.h"

void split2(int N, COMPLEX *X, COMPLEX *X1, COMPLEX *X2)
{
    int k;
    X1[0].real = X[0].real;
    X1[0].imag = 0;
    X2[0].real = X[0].imag;
    X2[0].imag = 0;
    X1[N/2].real = X[N/2].real;
    X1[N/2].imag = 0;
    X2[N/2].real = X[N/2].imag;
    X2[N/2].imag = 0;

    for (k=1; k<N/2; k++)
    {
        X1[k].real = (X[k].real + X[N-k].real)/2;
    }
\begin{align*}
X1[k].\text{imag} &= (X[k].\text{imag} - X[N-k].\text{imag})/2; \\
X2[k].\text{real} &= (X[k].\text{imag} + X[N-k].\text{imag})/2; \\
X2[k].\text{imag} &= (X[N-k].\text{real} - X[k].\text{real})/2; \\
X1[N-k].\text{real} &= X1[k].\text{real}; \\
X1[N-k].\text{imag} &= -X1[k].\text{imag}; \\
X2[N-k].\text{real} &= X2[k].\text{real}; \\
X2[N-k].\text{imag} &= -X2[k].\text{imag}; 
\end{align*}
Example 13. data2.c File

FILE
    data2.c - Sample data used in realdft2.c

*******************************************************************************
/* array of real-valued input sequence, x1(n) */
short x1[] = {255, 255, 255, 255, 255, 255, 255, 255,
          0, 0, 0, 0, 0, 0, 0, 0};

/* array of real-valued input sequence, x2(n) */
          0, 0, 0, 0, 0, 0, 0, 0};
Example 14. params2.h File

/****************************************************************************
FILE
    params2.h - This is the C header file for example real FFT implementations.
****************************************************************************/

#define NUMDATA 16     /* number of real data samples */
#define NUMPOINTS NUMDATA /* number of point in the DFT */
**Example 15.dft.c File**

```c
/****************************************************************************
FILE
    dft.c - This is the C source code for the direct implementation of the
Discrete Fourier Transform (DFT) algorithm.
****************************************************************************/

DESCRIPTION

This function computes the DFT of a N length complex-valued sequence. Note, N cannot exceed 1024 without modification to this code.

The N point DFT of a finite-duration sequence x(n) of length L<N is defined as

\[ X(k) = \sum_{n=0}^{N-1} x(n) \cdot \exp(-j2\pi k n/N) \quad k = 0, 1, 2, \ldots, N-1 \]

It is always helpful to express the above equation in its real and imaginary terms for implementation.

\[ \exp(-j2\pi n k/N) = \cos(2\pi n k/N) - jsin(2\pi n k/N) \rightarrow \text{several identities used here} \]

\[ e(jb) = \cos(b) + j \sin(b) \]
\[ e(-jb) = \cos(-b) + j \sin(-b) \]
\[ \cos(-b) = \cos(b) \quad \text{and} \quad \sin(-b) = -\sin(b) \]
\[ e(-jb) = \cos(b) - j \sin(b) \]

\[ X(k) = \sum_{n=0}^{N-1} \{[x_r(n) + j x_i(n)] \cdot [\cos(2\pi n k/N) - j\sin(2\pi n k/N)] \} \]
\[ k = 0, 1, 2, \ldots, N-1 \]

OR

\[ X_r(k) = \sum_{n=0}^{N-1} \{[x_r(n) \cdot \cos(2\pi n k/N)] + [x_i(n) \cdot \sin(2\pi n k/N)] \} \]
\[ k = 0, 1, 2, \ldots, N-1 \]

\[ X_i(k) = \sum_{n=0}^{N-1} \{[x_i(n) \cdot \cos(2\pi n k/N)] - [x_r(n) \cdot \sin(2\pi n k/N)] \} \]

****************************************************************************/

#include <math.h>
#include "params.h"
#include "params.h"

void dft(int N, COMPLEX *X)
{
    int n, k;
    double arg;
    int Xr[1024];
    int Xi[1024];
    short Wr, Wi;

    for(k=0; k<N; k++)
    {
        Xr[k] = 0;
    }
```
$X_i[k] = 0$;
for($n=0; n<N; n++$)
{
    \text{arg} = (2*PI*k*n)/N;
    Wr = (short)((double)32767.0 * cos(\text{arg}));
    Wi = (short)((double)32767.0 * sin(\text{arg}));
    Xr[k] = Xr[k] + X[n].real \times Wr + X[n].imag \times Wi;
    Xi[k] = Xi[k] + X[n].imag \times Wr - X[n].real \times Wi;
}

for ($k=0; k<N; k++$)
{
    X[k].real = (short)(Xr[k]>>15);
    X[k].imag = (short)(Xi[k]>>15);
}

Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences
Example 16.params.h File

/****************************************************************************
FILE
params.h - This is the C header file for example real FFT implementations.
******************************************************************************/

#define TRUE 1
#define FALSE 0
#define BE TRUE
#define LE FALSE
#define ENDIAN LE /* selects proper endianess. If
building code in Big Endian,
use BE, else use LE */

#define PI 3.141592653589793 /* definition of pi */

/* Some functions used in the example implementations use word loads which make
the code endianess dependent. Thus, one of the below definitions need to be
used depending on the endianess you are using to build your code */

/* BIG Endian */
#if ENDIAN == TRUE

typedef struct {
  short imag;
  short real;
} COMPLEX;
#else
/* LITTLE Endian */

typedef struct {
  short real;
  short imag;
} COMPLEX;
#endif

Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences
Example 17. vectors.asm

/***********************************************************************************/
/*  vectors.asm - reset vector assembly                                           */
/***********************************************************************************/

.def    RESET
.ref    _c_int00
.sect   "vectors"

RESET:
    mvk   .s2   _c_int00, B2
    mvkh  .s2   _c_int00, B2
    b     .s2   B2
    nop
    nop
    nop
    nop
    nop
Example 18. lnk.cmd

/***************************************************************
/*  lnk.cmd - example linker command file                      */
/***************************************************************

-c
-heap 0x2000
-stack 0x8000

MEMORY
{
    VECS: o = 00000000h  l=00200h /* reset & interrupt vectors*/
    IPRAM: o = 0000200h    l=0FE00h /* internal program memory */
    IDRAM: o = 80000000h  l=10000h /* internal data memory   */
}

SECTIONS
{
    vectors  >   VECS
    .text     >   IPRAM
    .tables   >   IDRAM
    .data     >   IDRAM
    .stack    >   IDRAM
    .bss      >   IDRAM
    .sysmem   >   IDRAM
    .cinit    >   IDRAM
    .const    >   IDRAM
    .cio      >   IDRAM
    .far      >   IDRAM
}
Optimized C Implementation of the DFT of Real Sequences

This appendix contains optimized C implementations of the efficient methods for performing the DFT of real-valued sequences outlined in this application report.

Implementation Notes

The following lists usage, assumption, and limitations of the code.

Data format

All data and state variables are 16-bit signed integers (shorts). In this example, the decimal point is assumed to be between bits 15 and 14, thus the Q15 data format. For complex data and variables, the real and imaginary components are both Q15 numbers. From this data format, you can see that this code was developed for a fixed-point processor.

Memory

Complex data is stored in memory in imaginary/real pairs. The imaginary component is stored in the most significant halfword (16 bits) and the real component is stored in the least significant halfword, unless otherwise noted.

Endianess

The code is presented and has been tested in little endian format. Some modification to the code is necessary for big endian format.

Overflow

No overflow protection or detection is performed.

File Description

<table>
<thead>
<tr>
<th>File</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>realdft3.c</td>
<td>DFT of a $2N$-point real sequence main program</td>
</tr>
<tr>
<td>realdft4.c</td>
<td>DFT of a two $N$-point real sequence main program</td>
</tr>
<tr>
<td>radix4.c</td>
<td>Radix-4 FFT C function</td>
</tr>
<tr>
<td>digit.c</td>
<td>Radix-4 digit reversal C function</td>
</tr>
<tr>
<td>digitgen.c</td>
<td>C function used to initialize digit reversal table used by the function in digit .c</td>
</tr>
<tr>
<td>splitgen.c</td>
<td>C function used to initialize the split tables used by the split1 routines</td>
</tr>
</tbody>
</table>
Example 19.realdft3.c File

/*********************************************/
FILE
  realdft3.c - C source for an example implementation of the DFT/IDFT
  of a 2N-point real sequences using one N-point complex DFT/IDFT.
*********************************************/

DESCRIPTION
This program is an example implementation of an efficient way of computing
the DFT/IDFT of a real-valued sequences.

In many applications, the input is a sequence of real numbers.
If this condition is taken into consideration, additional computational
savings can be achieved because the FFT of a real sequence has some
symmetrical properties. The DFT of a 2N-point real sequence can be
efficiently computed using a N-point complex DFT and some additional
computations.

The following steps are required in the computation of the FFT of
a real valued sequence using the split function:

1. Let g(n) be a 2N-point real sequence. From g(n), form the
the N-point complex valued sequence, x(n) = x1(n) + jx2(n),
where x1(n) = g(2n) and x2(n) = g(2n + 1).

2. Perform an N-point complex FFT on the complex valued sequence
x(n) -> X(k) = DFT{x(n)}. Note the FFT can be any DFT method,
such as radix-2, radix-4, mixed radix, direct implementation
of the DFT, etc. However, the DFT output must be in normal order.

3. The following additional computation are used to get G(k) from X(k)
Gr(k) = Xr(k)Ar(k) - Xi(k)Ai(k) + Xr(N-k)Br(k) + Xi(N-k)Bi(k)
k = 0, 1, ..., N-1
and X(N) = X(0)
Gi(k) = Xi(k)Ar(k) + Xr(k)Ai(k) + Xr(N-k)Bi(k) - Xi(N-k)Br(k)

Note, only N-points of the 2N-point sequence of G(k) is computed
in the above equations. Because the DFT of a real-sequence has
symmetric properties, we can easily compute the remaining N points
of G(k) with the following equations.
Gr(N) = Xr(0) - Xi(0)
Gi(N) = 0
Gr(2N-k) = Gr(k) k = 1, 2, ..., N-1
Gi(2N-k) = -Gi(k)

As you can see, the above equations assume that A(k) and B(k),
which are sine and cosine coefficients, are pre-computed. The
C-code can be used to initialize A(k) and B(k).

for(k=0; k<N; k++)
{  
  A[k].imag = (short)(16383.0*(-cos(2*PI/(double)(2*N)*(double)k)));
  A[k].real = (short)(16383.0*(1.0 - sin(2*PI/(double)(2*N)*(double)k)));
  B[k].imag = (short)(16383.0*(cos(2*PI/(double)(2*N)*(double)k)));
  B[k].real = (short)(16383.0*(1.0 + sin(2*PI/(double)(2*N)*(double)k)));
}

The following steps are required in the computation of the IFFT of
a complex valued frequency domain sequence, that was derived from
a real sequence:
1. Let $G(k)$ be a $2N$-point complex valued sequence derived from a real valued sequence $g(n)$. We want to get back $g(n)$ from $G(k) \rightarrow g(n) = \text{IDFT}(G(k))$. However, we want to apply the same techniques as we did with the forward FFT, use a $N$-point IFFT. This can be accomplished by the following equations.

$$X_r(k) = G_r(k)A_r(k) - G_i(k)A_i(k) + G_r(N-k)B_r(k) + G_i(N-k)B_i(k)$$
$$k = 0, 1, ..., N-1$$
and $G(N) = G(0)$

$$X_i(k) = G_i(k)A_r(k) + G_r(k)A_i(k) + G_r(N-k)B_i(k) - G_i(N-k)B_r(k)$$

2. Perform the $N$-point inverse DFT of $X(k) \rightarrow x(n) = x_1(n) + jx_2(n) = \text{IDFT}(X(k))$. Note, the IDFT can be any method, but must have an output that is in normal order.

3. $g(n)$ can then be found from $x(n)$.

$$g(2n) = x_1(n)$$
$$n = 0, 1, ..., N-1$$
$$g(2n+1) = x_2(n)$$

As you can see, the above equations can be used for both the forward and inverse FFTs, however, the pre-computed coefficients are slightly different. The following C-code can be used to initialize $IA(k)$ and $IB(k)$.

```c
for(k=0; k<N; k++)
{
    IA[k].imag = -(short)(16383.0*(-cos(2*PI/(double)(2*N)*(double)k)));
    IA[k].real = (short)(16383.0*(1.0 - sin(2*PI/(double)(2*N)*(double)k)));
    IB[k].imag = -(short)(16383.0*(cos(2*PI/(double)(2*N)*(double)k)));
    IB[k].real = (short)(16383.0*(1.0 + sin(2*PI/(double)(2*N)*(double)k)));
}
```

Note, $IA(k)$ is the complex conjugate of $A(k)$ and $IB(k)$ is the complex conjugate of $B(k)$.

*******************************************************************************************/

typedef struct { /* define the data type for the radix-4 twiddle factors */
    short imag;
    short real;
} COEFF;

#include "params1.h" /* header files with parameters */
#include "params.h"
#include "splatbl1.h" /* header file that contains tables used to generate
    the split tables */
#include "sinestbl.h" /* header file that contains the FFT twiddle factors */

#pragma DATA_ALIGN(x,64); /* radix-4 routine requires x to be
aligned to a 4*NUMPOINTS boundry */
COMPLEX x[NUMPOINTS+1]; /* array of complex DFT data */
extern short g[]; /* real-valued input sequence */

/* functions defined externally */
void FftSplitTableGen(int N, COMPLEX *W, COMPLEX *A, COMPLEX *B);
void R4DigitRevIndexTableGen(int, int *, unsigned short *, unsigned short *");
void split1(int, COMPLEX *, COMPLEX *, COMPLEX *, COMPLEX *");
void digit_reverse(int *, unsigned short *, unsigned short *, int);
void radix4(int, short[], short[]);
main()
{
    int n, k;
    COMPLEX A[NUMPOINTS]; /* array of complex A coefficients */
    COMPLEX B[NUMPOINTS]; /* array of complex B coefficients */
    COMPLEX IA[NUMPOINTS]; /* array of complex A* coefficients */
    COMPLEX IB[NUMPOINTS]; /* array of complex B* coefficients */
    COMPLEX G[2*NUMPOINTS]; /* array of complex DFT result */
    unsigned short IIndex[NUMPOINTS], JIndex[NUMPOINTS];
    int count;

    /* Initialize A, B, IA, and IB arrays */
    FftSplitTableGen(NUMPOINTS, W, A, B);

    /* Split tables for the IDFT are the complex conjugate of the split
    tables of the DFT */
    for(k=0; k<NUMPOINTS; k++)
    {
        IA[k].imag = -A[k].imag;
        IA[k].real = A[k].real;
        IB[k].imag = -B[k].imag;
        IB[k].real = B[k].real;
    }

    /* Initialize tables for FFT digit reversal function */
    R4DigitRevIndexTableGen(NUMPOINTS, &count, IIndex, JIndex);

    /* Forward DFT */
    /* From the 2N point real sequence, g(n), for the N-point complex sequence, x(n) */
    for (n=0; n<NUMPOINTS; n++)
    {
        x[n].imag = g[2*n + 1]; /* x2(n) = g(2n + 1) */
        x[n].real = g[2*n]; /* x1(n) = g(2n) */
    }

    /* Compute the DFT of x(n) to get X(k) -> X(k) = DFT{x(n)} */
    radix4(NUMPOINTS, (short *)x, (short *)W4);
    digit_reverse((int *)x, IIndex, JIndex, count);

    /* Because of the periodicity property of the DFT, we know that X(N+k)=X(k). */
    x[NUMPOINTS].real = x[0].real;
    x[NUMPOINTS].imag = x[0].imag;

    /* The split function performs the additional computations required to get
    G(k) from X(k). */
    split1(NUMPOINTS, x, A, B, G);

    /* Use complex conjugate symmetry properties to get the rest of G(k) */
    G[NUMPOINTS].real = x[0].real - x[0].imag;
    G[NUMPOINTS].imag = 0;
    for (k=1; k<NUMPOINTS; k++)
    {
        G[2*NUMPOINTS-k].real = G[k].real;
        G[2*NUMPOINTS-k].imag = -G[k].imag;
    
}
/* Inverse DFT - We now want to get back g(n). */

/* The split function performs the additional computations required to get X(k) from G(k). */
split1(NUMPOINTS, G, IA, IB, x);

/* Take the inverse DFT of X(k) to get x(n). Note the inverse DFT could be any IDFT implementation, such as an IFFT. */

/* The inverse DFT can be calculated by using the forward DFT algorithm directly by complex conjugation - x(n) = (1/N)\{DFT{X*(k)}\}*, where * is the complex conjugate operator. */

/* Compute the complex conjugate of X(k). */

for (k=0; k<NUMPOINTS; k++)
{
    x[k].imag = -x[k].imag; /* complex conjugate X(k) */
}

/* Compute the DFT of X*(k). */
radix4(NUMPOINTS, (short *)x, (short *)W4);
digit_reverse((int *)x, IIndex, JIndex, count);

/* Complex conjugate the output of the DFT and divide by N to get x(n). */

for (n=0; n<NUMPOINTS; n++)
{
    x[n].real = x[n].real/16;
    x[n].imag = (-x[n].imag)/16;
}

/* g(2n) = xr(n) and g(2n + 1) = xi(n) */

for (n=0; n<NUMPOINTS; n++)
{
    g[2*n] = x[n].real;
    g[2*n + 1] = x[n].imag;
}

return(0);

}
Example 20. realdft4.c File

/***********************************************************************************/
FILE
    realdft4.c - C source for an example implementation of the DFT/IDFT
    of two N-point real sequences using one N-point complex DFT/IDFT.
***********************************************************************************/

DESCRIPTION
This program is an example implementation of an efficient way of computing
the DFT/IDFT of two real-valued sequences.

Assume we have two real-valued sequences of length N - x1[n] and x2[n]. The
DFT of x1[n] and x2[n] can be computed with one complex-valued DFT of length
N, as shown above, by following this algorithm.

1. Form the complex-valued sequence x[n] from x1[n] and x2[n]
xr[n] = x1[n] and xi[n] = x2[n], 0,1, ..., N-1

   Note, if the sequences x1[n] and x2[n] are coming from another algorithm
   or a data acquisition driver, this step may be eliminated if these put the
   data in the complex-valued format correctly.

2. Compute X[k] = DFT{x[n]}
This can be the direct form DFT algorithm, or an FFT algorithm. If using an
FFT algorithm, make sure the output is in normal order - bit reversal is
performed.

3. Compute the following equations to get the DFTs of x1[n] and x2[n].
   X1r[0] = Xr[0]
   X1i[0] = 0
   X2r[0] = Xi[0]
   X2i[0] = 0
   X1r[N/2] = Xr[N/2]
   X1i[N/2] = 0
   X2r[N/2] = Xi[N/2]
   X2i[N/2] = 0
   for k = 1,2,3, ...., N/2-1
     X1r[k] = (Xr[k] + Xr[N-k])/2
     X1i[k] = (Xi[k] - Xi[N-k])/2
     X1r[N-k] = X1r[k]
     X1i[N-k] = X1i[k]
     X2r[k] = (Xi[k] + Xi[N-k])/2
     X2i[k] = (Xr[N-k] - Xr[k])/2
     X2r[N-k] = X2r[k]
     X2i[N-k] = X2i[k]

4. Form X[k] from X1[k] and X2[k]
   for k = 0,1, ..., N-1
     Xr[k] = X1r[k] - X2i[k]
     Xi[k] = X1i[k] + X2r[k]

5. Compute x[n] = IDFT{X[k]}
This can be the direct form IDFT algorithm, or an IFFT algorithm. If using an
IFFT algorithm, make sure the output is in normal order - bit reversal is
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

**Typedef**: struct { /* define the data type for the radix-4 twiddle factors */
  short imag;
  short real;
} COEFF;

#include "params2.h" /* include file with parameters */
#include "params.h" /* include file with parameters */
#include "sinestbl.h" /* header file that contains the FFT twiddle factors */

#pragma DATA_ALIGN(x,64); /* radix-4 routine requires x to be
aligned to a 4*NUMPOINTS boundary */

COMPLEX x[NUMPOINTS+1]; /* array of complex DFT data, X(k) */

extern short x1[];
extern short x2[];

void R4DigitRevIndexTableGen(int, int *, unsigned short *, unsigned short *);
void split2(int, COMPLEX *, COMPLEX *, COMPLEX *);
void digit_reverse(int *, unsigned short *, unsigned short *, int);
void radix4(int, short [], short []);

main()
{
  int n, k;
  COMPLEX X1[NUMDATA]; /* array of real-valued DFT output sequence, X1(k) */
  COMPLEX X2[NUMDATA]; /* array of real-valued DFT output sequence, X2(k) */
  unsigned short IIndex[NUMPOINTS], JIndex[NUMPOINTS];
  int count;

  /* Initialize tables for FFT digit reversal function */
  R4DigitRevIndexTableGen(NUMPOINTS, &count, IIndex, JIndex);

  /* Forward DFT */

  /* From the two N-point real sequences, x1(n) and x2(n), form the N-point complex
sequence, x(n) = x1(n) + jx2(n) */
  for (n=0; n<NUMDATA; n++)
  {
    x[n].real = x1[n];
    x[n].imag = x2[n];
  }

  /* Compute the DFT of x(n), X(k) = DFT{x(n)}. Note, the DFT can be any
DFT implementation such as FFTs. */
  radix4(NUMPOINTS, (short *)x, (short *)W4);
  digit_reverse((int *)x, IIndex, JIndex, count);

  /* Because of the periodicity property of the DFT, we know that X(N+k)=X(k). */
  x[NUMPOINTS].real = x[0].real;
  x[NUMPOINTS].imag = x[0].imag;
The split function performs the additional computations required to get $X_1(k)$ and $X_2(k)$ from $X(k)$. 

```c
split2(NUMPOINTS, x, X1, X2);
```

Inverse DFT - We now want to get back $x_1(n)$ and $x_2(n)$ from $X_1(k)$ and $X_2(k)$ using one complex DFT.

Recall that $x(n) = x_1(n) + jx_2(n)$. Since the DFT operator is linear, $X(k) = X_1(k) + jX_2(k)$. Thus we can express $X(k)$ in terms of $X_1(k)$ and $X_2(k)$.

```c
for (k=0; k<NUMPOINTS; k++)
{
    x[k].real = X1[k].real - X2[k].imag;
    x[k].imag = X1[k].imag + X2[k].real;
}
```

Take the inverse DFT of $X(k)$ to get $x(n)$. Note the inverse DFT could be any IDFT implementation, such as an IFFT.

The inverse DFT can be calculated by using the forward DFT algorithm directly by complex conjugation - $x(n) = (1/N)(DFT(X^*(k)))^*$, where $*$ is the complex conjugate operator.

Compute the complex conjugate of $X(k)$.

```c
for (k=0; k<NUMPOINTS; k++)
{
    x[k].imag = -x[k].imag;
}
```

Compute the DFT of $X^*(k)$.

```c
radix4(NUMPOINTS, (short *)x, (short *)W4);
digit_reverse((int *)x, IIndex, JIndex, count);
```

Complex conjugate the output of the DFT and divide by $N$ to get $x(n)$.

```c
for (n=0; n<NUMPOINTS; n++)
{
    x[n].real = x[n].real/16;
    x[n].imag = (-x[n].imag)/16;
}
```

$x_1(n)$ is the real part of $x(n)$, and $x_2(n)$ is the imaginary part of $x(n)$.

```c
for (n=0; n<NUMDATA; n++)
{
    x1[n] = x[n].real;
    x2[n] = x[n].imag;
}
```

return(0);
Example 21. radix4.c File

/***************************************************************************/
FILE
radix4.c - Radix-4 FFT function based on Burrus, Parks p 113
******************************************************************************/

void radix4(int n, short x[], short w[])
{
    int n1, n2, ie, ia1, ia2, ia3, i0, i1, i2, i3, j, k;
    short t, r1, r2, s1, s2, co1, co2, co3, si1, si2, si3;
    n2 = n;
    ie = 1;
    for (k = n; k > 1; k >>= 2) {
        n1 = n2;
        n2 >>= 2;
        ia1 = 0;
        for (j = 0; j < n2; j++) {
            ia2 = ia1 + ia1;
            ia3 = ia2 + ia2;
            co1 = w[ia1 * 2 + 1];
            si1 = w[ia1 * 2];
            co2 = w[ia2 * 2 + 1];
            si2 = w[ia2 * 2];
            co3 = w[ia3 * 2 + 1];
            si3 = w[ia3 * 2];
            ia1 = ia1 + ie;
            for (i0 = j; i0 < n; i0 += n1) {
                i1 = i0 + n2;
                i2 = i1 + n2;
                i3 = i2 + n2;
                r1 = x[2 * i0] + x[2 * i2];
                r2 = x[2 * i0] - x[2 * i2];
                t = x[2 * i1] + x[2 * i3];
                x[2 * i0] = r1 + t;
                r1 = r1 - t;
                s1 = x[2 * i0 + 1] + x[2 * i2 + 1];
                s2 = x[2 * i0 + 1] - x[2 * i2 + 1];
                t = x[2 * i1 + 1] + x[2 * i3 + 1];
                x[2 * i0 + 1] = s1 + t;
                s1 = s1 - t;
                x[2 * i2] = (r1 * co2 + s1 * si2) >> 15;
                x[2 * i2 + 1] = (s1 * co2 - r1 * si2) >> 15;
                t = x[2 * i1 + 1] - x[2 * i3 + 1];
                r1 = r2 + t;
                r2 = r2 - t;
                t = x[2 * i1] - x[2 * i3];
                s1 = s2 - t;
                s2 = s2 + t;
                x[2 * i1] = (r1 * co3 + s1 * si3) >> 15;
                x[2 * i1 + 1] = (s1 * co3 - r1 * si3) >> 15;
            }
        }
        ie <<= 2;
    }
}
Example 22. `digit.c` File

/**************************************************************************
FILE
digit.c - This is the C source code for a digit reversal function for
a radix-4 FFT
**************************************************************************/

void digit_reverse(int *yx, unsigned short *JIndex, unsigned short *IIndex, int count)
{
    int i;
    unsigned short I, J;
    int YXI, YXJ;
    for (i = 0; i<count; i++)
    {
        I = IIndex[i];
        J = JIndex[i];
        YXI = yx[I];
        YXJ = yx[J];
        yx[J] = YXI;
        yx[I] = YXJ;
    }
}
Example 23. digitgen.c File

/******************************************************************************
FILE
digitgen.c - This is the C source code for a function used to generate
index tables for a digit reversal function for a radix-4 FFT.
******************************************************************************/

void R4DigitRevIndexTableGen(int n, int *count, unsigned short *IIndex, unsigned short *JIndex)
{
    int j, n1, k, i;
    j = 1;
    n1 = n - 1;
    *count = 0;
    for(i=1; i<=n1; i++)
    {
        if(i < j)
        {
            IIndex[*count] = (unsigned short)(i-1);
            JIndex[*count] = (unsigned short)(j-1);
            *count = *count + 1;
        }
        k = n >> 2;
        while(k*3 < j)
        {
            j = j - k*3;
            k = k >> 2;
        }
        j = j + k;
    }
}
Example 24. splitgen.c File

/***************************************************************************/
FILE
    splitgen.c - This is the C source code for a function used to generate
tables for split routine used to efficiently compute the DFT of a 2N-point
real-valued sequence.
***************************************************************************/

#include "params.h"

void FftSplitTableGen(int N, COMPLEX *W, COMPLEX *A, COMPLEX *B) {
    int k;

    for(k=0; k<N/2; k++) {
        A[k].real = 16383 - W[k].imag;
        A[k].imag = -W[k].real;
        A[k + N/2].real = 16383 - W[k].real;
        A[k + N/2].imag = W[k].imag;
        B[k].real = 16383 + W[k].imag;
        B[k].imag = W[k].real;
        B[k + N/2].real = 16383 + W[k].real;
        B[k + N/2].imag = -W[k].imag;
    }
}
Appendix D. Optimized C Callable ’C62xx Assembly Language Functions Used to Implement the DFT of Real Sequences

This appendix contains optimized C callable ’C62xx assembly language functions used to implement the DFT of real sequences.

Implementation Notes

The following lists usage, assumption, and limitations of the code.

**Data format**
All data and state variables are 16-bit signed integers (short’s). In this example, the decimal point is assumed to be between bits 15 and 14, thus the Q15 data format. For complex data and variables, the real and imaginary components are both Q15 numbers. From this data format, you can see that this code was developed for a fixed-point processor.

**Memory**
Complex data is stored in memory in imaginary/real pairs. The imaginary component is stored in the most significant halfword (16 bits) and the real component is stored in the least significant halfword, unless otherwise noted.

**Endianess**
The code is presented and has been tested in little endian format. Some modification to the code is necessary for big endian format.

**Overflow**
No overflow protection or detection is performed.

**File** **Description**

<table>
<thead>
<tr>
<th>File</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>split1.asm</td>
<td>C callable ’C62xx assembly version of the split function for the DFT of a 2N-point real sequence</td>
</tr>
<tr>
<td>split2.asm</td>
<td>C callable ’C62xx assembly version of the split function for the DFT of a 2N-point real sequences.</td>
</tr>
<tr>
<td>radix4.asm</td>
<td>Radix-4 FFT C callable ’C62xx assembly function.</td>
</tr>
<tr>
<td>digit.asm</td>
<td>Radix-4 digit reversal C callable ’C62xx assembly function.</td>
</tr>
</tbody>
</table>
Example 25. split1.asm File

===============================================================================
* TEXAS INSTRUMENTS, INC.
* Real FFT/IFFT split operation
* Revision Date: 5/15/97
* USAGE This routine is C Callable and can be called as:
  * void split1(int N, COMPLEX *X, COMPLEX *A, COMPLEX *B, COMPLEX *G)
  * N = 1/2 the number of samples of the real valued sequence
  * X = pointer to complex input array
  * A = pointer to complex coefficients
  * B = pointer to complex coefficients
  * G = pointer to complex output array
  * If routine is not to be used as a C callable function
  * then all instructions relating to stack should be removed.
  * Refer to comments of individual instructions. You will also
  * need to initialize values for all of the values passed as these
  * are assumed to be in registers as defined by the calling
  * convention of the compiler, (refer to the C compiler reference
  * guide).
* C Code This is the C equivalent of the Assembly Code without
  * restrictions. Note that the assembly code is hand optimized and
  * restrictions may apply.
  * One small, but important note. The split functions uses word loads
  * to read imaginary/real pairs from memory. Because of this, some C
  * definitions may need to be endianess dependent. Below are the
  * type definitions for COMPLEX for both big and little endian. Also,
  * the split function as shown below is written for big endian. See
  * comments in the code to see how to modify if little endian is desired.

LITTLE ENDIAN            BIG ENDIAN
typedef struct {         typedef struct {
  short real;            short real;
  short imag;            short imag;
} COMPLEX; } COMPLEX;

void split(int N, COMPLEX *X, COMPLEX *A, COMPLEX *B, COMPLEX *G)
{
  int k;
  int Tr, Ti;
  for (k=0; k<N; k++)
    {
      Tr = (int)X[k].real * (int)A[k].real -
          (int)X[k].imag * (int)A[k].imag +
          (int)X[N-k].real * (int)B[k].real +
          (int)X[N-k].imag * (int)B[k].imag;
      G[k].real = (short)(Tr>>15);
      Ti = (int)X[k].imag * (int)A[k].real +
          (int)X[k].real * (int)A[k].imag -
          (int)X[N-k].real * (int)B[k].imag -
          (int)X[N-k].imag * (int)B[k].real;
      G[k].imag = (short)(Ti>>15);
    }
}

DESCRIPTION
  In many applications, the input is a sequence of real numbers.
  If this condition is taken into consideration, additional computational
  savings can be achieved because the FFT of a real sequence has some
  symmetrical properties. The DFT of a 2N-point real sequence can be
SPRA291

76 Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

efficiently computed using a N-point complex DFT and some additional computations which have been implemented in the split function. Note this split function can be used in the computation of FFTs and IFFTs.

The following steps are required in the computation of the FFT of a real valued sequence using the split function:

1. Let \( g(n) \) be a 2N-point real sequence. From \( g(n) \), form the N-point complex valued sequence, \( x(n) = x_1(n) + jx_2(n) \), where \( x_1(n) = g(2n) \) and \( x_2(n) = g(2n + 1) \).

2. Perform an N-point complex FFT on the complex valued sequence \( x(n) \rightarrow X(k) = DFT\{x(n)\} \). Note the FFT can be any DFT method, such as radix-2, radix-4, mixed radix, direct implementation of the DFT, etc. However, the DFT output must be in normal order.

3. The following additional computations are used to get \( G(k) \) from \( X(k) \) and are implemented by the split function.

\[
\begin{align*}
Gr(k) &= X_r(k)A_r(k) - X_i(k)A_i(k) + X_r(N-k)B_r(k) + X_i(N-k)B_i(k) \\
Gi(k) &= X_i(k)A_r(k) + X_r(k)A_i(k) + X_r(N-k)B_i(k) - X_i(N-k)B_r(k)
\end{align*}
\]

Note, only N-points of the 2N-point sequence of \( G(k) \) is computed in the above equations. Because the DFT of a real-sequence has symmetric properties, we can easily compute the remaining N points of \( G(k) \) with the following equations.

\[
\begin{align*}
Gr(N) &= Gr(0) - Gi(0) \\
Gi(N) &= 0 \\
Gr(2N-k) &= Gr(k) \quad k = 1, 2, \ldots, N-1 \\
Gi(2N-k) &= -Gi(k)
\end{align*}
\]

As you can see, the split function assumes that \( A(k) \) and \( B(k) \), which are sine and cosine coefficients, are pre-computed. The C-code can be used to initialize \( A(k) \) and \( B(k) \).

\[
\text{for(k=0; k<N; k++)} \\
\quad \{ \\
\quad \quad A[k].\text{imag} = (\text{short})(16383.0*(-\cos(2*\pi/(\text{double})(2*N)*(\text{double)k}))); \\
\quad \quad A[k].\text{real} = (\text{short})(16383.0*(1.0 - \sin(2*\pi/(\text{double})(2*N)*(\text{double)k}))); \\
\quad \quad B[k].\text{imag} = (\text{short})(16383.0*(\cos(2*\pi/(\text{double})(2*N)*(\text{double)k}))); \\
\quad \quad B[k].\text{real} = (\text{short})(16383.0*(1.0 + \sin(2*\pi/(\text{double})(2*N)*(\text{double)k}))); \\
\quad \}
\]

The following steps are required in the computation of the IFFT of a complex valued frequency domain sequence, that was derived from a real sequence, using the split function:

1. Let \( G(k) \) be a 2N-point complex valued sequence derived from a real valued sequence \( g(n) \). We want to get back \( g(n) \) from \( G(k) \rightarrow g(n) = IDFT\{G(k)\} \). However, we want to apply the same techniques as we did with the forward FFT, use a N-point IFFT. This can be accomplished by the following equations.

\[
\begin{align*}
X_r(k) &= Gr(k)I_Ar(k) - Gi(k)I_Ai(k) + Gr(N-k)I_Br(k) + Gi(N-k)I_Bi(k) \\
X_i(k) &= Gi(k)I_Ar(k) + Gr(k)I_Ai(k) + Gr(N-k)I_Bi(k) - Gi(N-k)I_Br(k)
\end{align*}
\]

2. Perform the N-point inverse DFT of \( X(k) \) \( \rightarrow x(n) = x_1(n) + jx_2(n) = IDFT\{X(k)\} \). Note, the IDFT can be any method, but must have an output that is in normal order.

3. \( g(n) \) can then be found from \( x(n) \).

\[
\begin{align*}
g(2n) &= x_1(n) \\
g(2n+1) &= x_2(n)
\end{align*}
\]

As you can see, the split function can be used for both the forward and inverse FFTs, however, the pre-computed coefficients are slightly different. The following C-code can be used to initialize \( I_A(k) \) and \( I_B(k) \).

\[
\text{for(k=0; k<N; k++)}
\]
{  
    IA[k].imag = -(short)(16383.0*(-cos(2*PI/(double)(2*N)*(double)k)));
    IA[k].real = (short)(16383.0*(1.0 - sin(2*PI/(double)(2*N)*(double)k)));
    IB[k].imag = -(short)(16383.0*(cos(2*PI/(double)(2*N)*(double)k)));
    IB[k].real = (short)(16383.0*(1.0 + sin(2*PI/(double)(2*N)*(double)k)));
}

Note, IA(k) is the complex conjugate of A(k) and IB(k) is the complex
conjugate of B(k).

TECHNIQUES
32-bit loads are used to load two 16-bit loads.

ASSUMPTIONS
A, B, X, and G are stored as imaginary/real pairs.

Big endian is used. If little endian is desired, modification to
the code is required. See comments in the code for which instructions
require modification for little endian use.

MEMORY NOTE
A, B, X and G arrays should be aligned to word boundaries. Also,
A and B should be aligned such that they do not generate a memory
hit with X.

A, B, X, and G are all complex data and are required to be stored
as imaginary/real pairs in memory, irregardless of endianess. In other
words, a load word from any of these arrays should result in the imaginary
component in the upper 16 bits of a register and the real component in the
lower 16 bits.

CYCLES 4*N + 32

===============================================================================
N .set a4 ; argument 1 - number of points in the FFT
XPtr .set b4 ; argument 2 - pointer to complex data
APtr .set a6 ; argument 3 - pointer to A complex coefficients
BPtr .set b6 ; argument 4 - pointer to B complex coefficients
GPtr .set a8 ; argument 5 - pointer to output buffer
aI_aR .set a0 ; coefficient value loaded from APtr
XNPtr .set a1 ; pointer to the bottom of the data buffer
            ; this pointer gets decremented through the loop
x2I_x2R .set a2 ; data value loaded from XNPtr
xRaR .set a3 ; product
x1I .set a5
x2RbR .set a7
x2IbI .set a9
re1 .set a10
re2 .set a11
real .set a12
CNT .set b0 ; counter for looping
x1I_xR .set b1
b1_br .set b2
x1aR .set b5
xRai .set b7
x2IbR .set b8
x2RbI .set b9
im1 .set b10
im2 .set b11
imag .set b12

.global _split1
_split1:

Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences
sub .d2 B15,24,B15 ; allocate space on the stack
stw .d2 A10,*B15++[1] ; push A10 onto the stack
stw .d2 A11,*B15++[1] ; push A11 onto the stack
stw .d2 A12,*B15++[1] ; push A12 onto the stack
stw .d2 B10,*B15++[1] ; push B10 onto the stack
|| sub .l2x N,1,CNT ; initialize loop count register
|| stw .d2 B11,*B15++[1] ; push B11 onto the stack
|| shl .s1 N,2,N ; calculate offset to initialize
|| ; a pointer to the bottom of the
|| ; input data buffer
|| stw .d2 B12,*B15++[1] ; push B12 onto the stack
|| add .l1x N,XPtr,XNPtr ; XNPtr -> yx[N]
;
; Because there are delay slots in loads, we will begin by
; SW pipelining the split operations - in other words, while
; we are finishing the current loop iteration, we will be
; beginning the next.
ldw .d1 *APtr++[1],aI_aR ; load a coefficient pointed by APtr
ldw .d2 *XPtr++[1],xI_xR ; load a data value pointed by XPtr
nop ; fill a delay slot
ldw .d1 *XPTR+-[1],x2I_x2R ; load a data value pointed by XNPtr
ldw .d2 *BPtr++[1],bI_bR ; load a coefficient pointed by BPtr
nop ; fill a delay slot
ldw .d1 *APTR++[1],aI_aR ; load the next value pointed by APtr
|| ; (note, it will not overwrite the
|| ; the current value of aI_aR until
|| ; 4 delay slots later).
ldw .d2 *XPTR++[1],xI_xR ; load the next value pointed by XPtr
;
; for performing the multiplies, we take advantage of the feature
; feature that allows you to choose the operands from either the upper
; or lower halves of the register
mpy .m1x xI_xR,aI_aR,xRaR ; xRaR = xR * aR - mpy lower * lower
mpyh .m2x xI_xR,aI_aR,xRaR ; xRaR = xR * aR - mpy lower * lower
mpyh .m2x xI_xR,aI_aR,xIaR ; xIaR = xI * aR - mpy upper * lower
mpy .m1x xI_xR,aI_aR,xIaI ; xIaI = xI * aI - mpy upper * upper
ldw .d1 *XNPtr+-[1],x2I_x2R ; load a data value pointed by XNPtr
ldw .d2 *BPtr++[1],bI_bR ; load a coefficient pointed by BPtr
mpy .m2x x2I_x2R,bI_bR,x2RbR ; x2RbR = x2R * bR - mpy lower * lower
mpyh .m2x x2I_x2R,bI_bR,x2RbR ; x2RbR = x2R * bR - mpy lower * lower
mpyhl .m2x x2I_x2R,bI_bR,x2RbI ; x2RbI = x2R * bI - mpy lower * upper
mpyh .m1x x2I_x2R,bI_bR,xIaR ; xIaR = xI * aR - mpy upper * upper
sub .l1 xRaR,xIaR,reI ; reI = xRaR - xIaI
add .l2 xRai,xIaR,imI ; imI = xRai + xIaR
ldw .d1 *APTR++[1],aI_aR ; 3rd load of aI_aR
ldw .d2 *XPTR++[1],xI_xR ; 3rd load of xI_xR
;
; the second loads of xI_xR and aI_aR are now available, thus we can use
; them to begin the 2nd iteration of X's and A's multiplies
mpy .m1 xI_xR,aI_aR,xRaR ; xRaR = xR * aR - mpy lower * lower
mpyh .m2 xI_xR,aI_aR,xIaR ; xIaR = xI * aR - mpy upper * lower
mpyhl .m2 xI_xR,aI_aR,xIaI ; xIaI = xI * aI - mpy upper * upper
add .l1 x2RbR,x2IbR,re2 ; re2 = x2RbR + x2IbI
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

```
| sub .d2 x2RbI,x2IbR,im2 | ; im2 = x2RbI - x2IbR |
| ldw .d2 *BPtr++[1],bI_bR | ; 3rd load of bI_bR |

; the second loads of x2I_x2R and bI_bR are now available, thus we can use
; them to begin the 2nd iteration of X2's and B's multiplies

| mpy .m1 x2I_x2R,bI_bR,x2RbR | ; x2RbR = x2R * bR - mpy lower * lower |
| mpyhl .m2x x2I_x2R,bI_bR,x2IbR | ; x2IbR = x2I * bR - mpy upper * lower |
| add .d1 rel,re1,real | ; real = rel + re1 |
| add .d2 im1,im2,imag | ; imag = im1 + im2 |
| b .s2 LOOP | ; branch to LOOP - note, this is the |
| ; branch for the first time through |
| ; the loop, because of this, we only |
| ; need to do count-1 branches to LOOP |
| ; within LOOP |

| mpyhl .m2 x2I_x2R,bI_bR,x2RbI | ; x2RbI = x2R * bI - mpy lower * upper |
| mpyh .m1x x2I_x2R,bI_bR,x2IbR | ; x2IbR = x2I * bI - mpy upper * upper |
| sub .d1 xRaR,xIaI,rel | ; rel = xRaR - xIaI |
| shr .s1 real,15,real | ; real = real >> 15 |
| shr .s2 imag,15,imag | ; imag = imag >> 15 |
| | LOOP: ; this loop is executed N times |
| mpy .m1 x1_xR,aI_aR,xRaR | ; xRaR = xR * aR - mpy lower * lower |
| mpyhl .m2x x1_xR,aI_aR,xIaR | ; xIaR = xI * aR - mpy upper * lower |
| sth .d1 imag,*GPtr++[1] | ; store imag in output buffer |
| ; CAUTION - Big Endian specific code |
| ; if Little Endian is desired, |
| ; replace this line with: |
| sth .d1 real,*GPtr++[1] | ; store real in output buffer |
| ; CAUTION - Big Endian specific code |
| ; if Little Endian is desired, |
| ; replace this line with: |
| sth .d1 imag,*GPtr++[1] | ; store imag in output buffer |
| ; CAUTION - Big Endian specific code |
| ; if Little Endian is desired, |
| ; replace this line with: |
| sth .d1 real,*GPtr++[1] | ; store real in output buffer |

| [CNT] sub .d2 CNT,1,CNT | ; if (CNT != 0), CNT = CNT - 1 |

| mpyh .m1x x2I_x2R,bI_bR,x2RbR | ; x2RbR = x2R * bR - mpy lower * lower |
| mpyh .m2x x2I_x2R,bI_bR,x2IbR | ; x2IbR = x2I * bR - mpy upper * lower |
| add .d1 xRaI,xIaR,im1 | ; im1 = xRaI + xIaR |
| | [CNT] b .s2 LOOP | ; if (CNT != 0), branch to LOOP |
| ; end of LOOP |

| mpyhl .m2x x2I_x2R,bI_bR,x2RbI | ; x2RbI = x2R * bI - mpy lower * upper |
| mpyh .m1x x2I_x2R,bI_bR,x2IbI | ; x2IbI = x2I * bI - mpy upper * upper |
| sub .d1 xRaI,xIaI,rel | ; rel = xRaI - xIaI |
| shr .s1 real,15,real | ; real = real >> 15 |
| shr .s2 imag,15,imag | ; imag = imag >> 15 |
| | ldw .d1 *XPtr++[1],aI_aR | ; next load of aI_aR |
| ldw .d2 *XPtr++[1],x1_xR | ; next load of x1_xR |

ldw .d2 *--B15[1],B12 | ; pop B12 from the stack |
ldw .d2 *--B15[1],B11 | ; pop B11 from the stack |
```
ldw .d2 *--B15[1],B10; pop B10 from the stack
ldw .d2 *--B15[1],A12; pop A12 from the stack
ldw .d2 *--B15[1],A11; pop A11 from the stack
ldw .d2 *--B15[1],A10; pop A10 from the stack
b .s2 B3; function return
add .d2 B15,24,B15; de-allocate space from the stack
nop 4; fill delay slots
Example 26.split2.asm File

*===============================================================================
* TEXAS INSTRUMENTS, INC.
* Real FFT/IFFT split operation
* Revision Date: 6/4/97
* USAGE This routine is C Callable and can be called as:
* 
* void split2(int N, COMPLEX *X, COMPLEX *X1, COMPLEX *X2)
* 
* N = the number of samples of each real valued sequence
* X = pointer to complex input array
* X1 = pointer to complex array DFT result of sequence 1
* X2 = pointer to complex array DFT result of sequence 2
* 
* If routine is not to be used as a C call able function
* then all instructions relating to stack should be removed.
* Refer to comments of individual instructions. You will also
* need to initialize values for all of the values passed as these
* are assumed to be in registers as defined by the calling
* convention of the compiler, (refer to the C compiler reference
* guide).
* 
* C Code This is the C equivalent of the Assembly Code without
* restrictions. Note that the assembly code is hand optimized and
* restrictions may apply.
* 
* One small, but important note. The split functions uses word loads
* to read imaginary/real pairs from memory. Because of this, some C
* definitions may need to be endianess dependent. Also,
* the split function as shown below is written for little endian.
* 
* typedef struct {
*     short real;
*     short imag;
* } COMPLEX;
* 
* void split2(int N, COMPLEX *X, COMPLEX *X1, COMPLEX *X2)
* {
*     int k;
*     X1[0].real = X[0].real;
*     X1[0].imag = 0;
*     X2[0].real = X[0].imag;
*     X2[0].imag = 0;
*     X1[N/2].real = X[N/2].real;
*     X1[N/2].imag = 0;
*     X2[N/2].real = X[N/2].imag;
*     X2[N/2].imag = 0;
*     
*     for (k=1; k<N/2; k++)
*     {
*         X1[k].real = (X[k].real + X[N-k].real)/2;
*         X1[k].imag = (X[k].imag - X[N-k].imag)/2;
*         X2[k].real = (X[k].imag + X[N-k].imag)/2;
*         X2[k].imag = (X[N-k].real - X[k].real)/2;
*         X1[N-k].real = X1[k].real;
*         X1[N-k].imag = -X1[k].imag;
*         X2[N-k].real = X2[k].real;
*         X2[N-k].imag = -X2[k].imag;
*     }
* }
* 
* }
DESCRIPTION

In many applications, the input is a sequence of real numbers. If this condition is taken into consideration, additional computational savings can be achieved because the FFT of a real sequence has some symmetrical properties. The DFT of a two N-point real sequence can be efficiently computed using one N-point complex DFT and some additional computations which have been implemented in this split function. Note this split function can be used in the computation of FFTs and IFFTs.

The following steps are required in the computation of the FFT of two real valued sequence using the split function:

Assume we have two real-valued sequences of length N - x1[n] and x2[n]. The DFT of x1[n] and x2[n] can be computed with one complex-valued DFT of length N, as shown above, by following this algorithm.

1. Form the complex-valued sequence x[n] from x1[n] and x2[n]
   \[ x_r[n] = x_1[n] \quad \text{and} \quad x_i[n] = x_2[n], \quad 0,1, \ldots, N-1 \]
   Note, if the sequences x1[n] and x2[n] are coming from another algorithm or a data acquisition driver, this step may be eliminated if these put the data in the complex-valued format correctly.

2. Compute \( X[k] = \text{DFT}\{x[n]\} \)
   This can be the direct form DFT algorithm, or an FFT algorithm. If using an FFT algorithm, make sure the output is in normal order - bit reversal is performed.

3. Compute the following equations to get the DFTs of x1[n] and x2[n]. These are the equations that this file implements.
   \[
   \begin{align*}
   X_1r[0] &= X_r[0] \\
   X_1i[0] &= 0 \\
   X_2r[0] &= X_i[0] \\
   X_2i[0] &= 0 \\
   X_1r[N/2] &= X_r[N/2] \\
   X_1i[N/2] &= 0 \\
   X_2r[N/2] &= X_i[N/2] \\
   X_2i[N/2] &= 0 \\
   \end{align*}
   \]
   for \( k = 1,2,3, \ldots, N/2-1 \)
   \[
   \begin{align*}
   X_1r[k] &= (X_r[k] + X_r[N-k])/2 \\
   X_1i[k] &= (X_i[k] - X_i[N-k])/2 \\
   X_1r[N-k] &= X_1r[k] \\
   X_1i[N-k] &= X_1i[k] \\
   X_2r[k] &= (X_i[k] + X_i[N-k])/2 \\
   X_2i[k] &= (X_r[N-k] - X_r[k])/2 \\
   X_2r[N-k] &= X_2r[k] \\
   X_2i[N-k] &= X_2i[k] \\
   \end{align*}
   \]

4. Form X[k] from X1[k] and X2[k]
   for \( k = 0,1, \ldots, N-1 \)
   \[
   \begin{align*}
   x_r[k] &= X_1r[k] - X_2i[k] \\
   x_i[k] &= X_1i[k] + X_2r[k] \\
   \end{align*}
   \]

5. Compute \( x[n] = \text{IDFT}\{X[k]\} \)
   This can be the direct form IDFT algorithm, or an IFFT algorithm. If using an IFFT algorithm, make sure the output is in normal order - bit reversal is performed.
TECHNIQUES
32-bit loads are used to load two 16-bit loads.

ASSUMPTIONS
X, X1, and X2 are stored as imaginary/real pairs.
Little endian is used. If little endian is desired, modification to the code is required.

MEMORY NOTE
X must be aligned to a 32-bit boundary
CYCLES 5*(N/2-1) + 29

---

N .set a4
XPtr .set b4
X1Ptr .set a6
X2Ptr .set b6
CNT .set b0
XNmknPtr .set a3
N4 .set a0
Xixr .set b2
XNiXNr .set a2
X1NmknPtr .set a1
X2NmknPtr .set b1
X2rX1r .set a8
X1iX2i .set b8
Xir .set a9
Xil .set b9
X2r .set a10
X2i .set b10
X1Nr .set a12
X1Ni .set b12
X2Nr .set a13
X2Ni .set b13
nullA .set a14
nullB .set b5

.global _split2

_split2:
subaw .d2 B15,10,B15 ; allocate space on the stack
ldh .d2 *XPtr,X1r ; X1r = Xr[0]
|| add .l B15,4,A15 ; A15 points to the stack as well
|| ldh .d2 *XPtr[1],X2r ; X2r = Xi[0]
|| stw .d1 A10,*A15++[2] ; push A10 onto the stack
|| stw .d2 B10,*B15++[2] ; push B10 onto the stack
|| stw .d1 A11,*A15++[2] ; push A11 onto the stack
|| stw .d2 B11,*B15++[2] ; push B11 onto the stack
|| stw .d1 A12,*A15++[2] ; push A12 onto the stack
|| stw .d2 B12,*B15++[2] ; push B12 onto the stack
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

```assembly
stw .d1 A13,*A15++[2] ; push A13 onto the stack
stw .d1 A14,*A15++[2] ; push A14 onto the stack
stw .d2 B14,*B15++[2] ; push B14 onto the stack
stw .d1 X1r, *X1Ptr ; X1r[0]=Xr[0]
shl .s1 N, 2, N4 ; N4 = 4*N
sub .l1 N4, 4, N4 ; N4 = N4 - 1
sth .d1 nullA, **X1Ptr[1] ; X1i[0]=0
sth .d2 nullB, **X2Ptr[1] ; X2i[0]=0
add .l1 N4, XPtr, XNmkPtr ; XNmkPtr -> X[N-1]
add .l2 XPtr, 4, XPtr ; XPtr -> X[1]
ldw .d2 *XPtr++[1], XiXr ; load X[k].real and X2[k].imag
ldw .d1 *XNmkPtr--[1], XNiXNr ; load X[N-k].real and X[N-k].imag
shr .s2x N, 1, CNT ; CNT = N/2
sub .s2 CNT, 1, CNT ; CNT = N/2 - 1
add .l2 X1Ptr, 4, X1Ptr ; X1Ptr -> X1[1]
add .l2 X2Ptr, 4, X2Ptr ; X2Ptr -> X2[1]
add2 .s1x X1r, XNiXNr, X2rX1r ; X2[k].real = X[k].imag + X[N-k].imag
    (upper 16 bits)
    X1[k].real = X[k].real + X[N-k].real
    (lower 16 bits)
|| sub2 .s2x X1r, XNiXNr, X1iX2i ; X1[k].imag = X[k].imag - X[N-k].imag
    (upper 16 bits)
    X2[k].imag = X[k].real - X[N-k].real
    (lower 16 bits)
    upper 16 bits are don't cares
shr .s1 X2rX1r, 17, X2r ; X2r = X2rX1r>>17
|| sub2 .s2 nullB, X1iX2i, X2i ; X2[k].imag = X[N-k].real - X[k].real
    (lower 16 bits)
    X1i = X1iX2i>>17
shr .s2 X1iX2i, 17, X1i ; X1i = X1iX2i>>17
|| b .s1 LOOP ; branch for the first time through loop
LOOP:
ext .s1 X2rX1r, 16,17, X1r ; X1r[0]=Xr[0]
ext .s2 X2i, 16,17, X2i ; X2[k].imag = X2[k].imag/2
sth .d1 X1i, **X1Ptr[1] ; store X1[k].imag
sth .d2 X2r, **X2Ptr++[1] ; store X2[k].real
mv .l1 X1r, X1Nr ; X1[N-k].real = X1[k].real
mv .s1 X2r, X2Nr ; X2[N-k].real = X2[k].real
sub .l2 nullB, X1i, X1Ni ; X1[N-k].imag = -X1[k].imag
sub .s2 nullB, X2i, X2Ni ; X2[N-k].imag = -X2[k].imag
sth .d2 X2i, **X2Ptr++[1] ; store X2[k].imag
sth .d1 X1r, *X1Ptr++[2] ; store X1[k].real
add2 .s1x X1r, XNiXNr, X2rX1r ; X2[k].real = X[k].imag + X[N-k].imag
```

84

Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

`; (upper 16 bits)
; X1[k].real = X[k].real + X[N-k].real
; (lower 16 bits)

|| sub2 .s2x XiXr, XNiXNr, X1iX2i ; X1[k].imag = X[k].imag - X[N-k].imag
|| ldw .d2 *XPtr++[1], XiXr ; load X[k].real and X2[k].imag
|| ldw .d1 *XNmkPtr--[-1], XNiXNr ; load X[N-k].real and X[N-k].imag

shr .s1 X2rX1r, 17, X2r ; X2[k].real = (X[k].imag + X[N-k].imag)/2
|| sub2 .s2 nullB, X1iX2i, X2i ; X2[k].imag = X[N-k].real - X[k].real
; (lower 16 bits)
|| sth .d1 X1Ni, **X1NmkPtr[1] ; store X1[N-k].imag
|| sth .d2 X2Nr, **X2NmkPtr--[-2] ; store X2[N-k].real
|| [CNT] sub .l2 CNT, 1, CNT ; decrement loop counter

shr .s2 X1iX2i, 17, X1i ; X1[k].imag = (X[k].imag + X[N-k].imag)/2
|| sth .d1 X1Ni, **X1NmkPtr--[-2] ; store X1[N-k].real
|| sth .d2 X2Ni, **X2NmkPtr--[-3] ; store X2[N-k].imag
|| [CNT] b .s1 LOOP ; conditional branch

; LOOP END

|| ldw .d1 "--A15[2],A14 ; pop A14 from the stack
|| ldw .d2 "--B15[2],B14 ; pop B14 from the stack
|| ldw .d1 "--A15[2],A13 ; pop A13 from the stack
|| ldw .d2 "--B15[2],B13 ; pop B13 from the stack
|| ldw .d2 "--B15[2],B12 ; pop B12 from the stack
|| ldw .d1 "--A15[2],A12 ; pop A12 from the stack
|| ldw .d2 "--A15[2],A11 ; pop A11 from the stack
|| ldw .d2 "--B15[2],B11 ; pop B11 from the stack
|| b .s2 B3 ; function return
|| ldw .d2 "--B15[2],B10 ; pop B10 from the stack
|| ldw .d1 "--A15[2],A10 ; pop A10 from the stack
|| addaw .d2 B15,10,B15 ; deallocate space from the stack
|| nop 3 ; fill delay slots
Example 27. radix4.asm File

*******************************************************************************
T</noindent>EXAS INSTRUMENTS INC.

COMPLEX FFT (Radix 4)

Revision Data: 04/28/97

USAGE This routine is C callable and can the called as

```c
void radix4(int n, short x[], short w[])
```

- **n**: FFT size (power of 4) (input)
- **x[]**: input and output sequences (dim-n) (input/output)
- **w[]**: FFT coefficients (dim-n) (input)

If the routine is not to be used as a C callable function,
then all instructions relating to dummy should be removed.
Refer to comments of individual instructions. You will also
need to initialize values for all the values passed as these
are assumed to be in registers as defined by the calling
convention of the compiler, (refer to the C compiler reference
guide.)

C CODE

This is the C equivalent of the Assembly Code without the
assumptions listed below. Note that the assembly code is hand
optimized and assumptions apply.

SOURCE: Burrus, Parks p.113

```c
void radix4(int n, short x[], short w[])
{
    int n1, n2, ie, ia1, ia2, ia3, i0, i1, i2, i3, j, k;
    short t, r1, r2, s1, s2, co1, co2, co3, si1, si2, si3;
    n2 = n;
    ie = 1;
    for (k = n; k > 1; k >>= 2) {
        n1 = n2;
        n2 >>= 2;
        ia1 = 0;
        for (j = 0; j < n2; j++) {
            ia2 = ia1 + ia1;
            ia3 = ia2 + ia1;
            co1 = w[ia1 * 2 + 1];
            si1 = w[ia1 * 2];
            co2 = w[ia2 * 2 + 1];
            si2 = w[ia2 * 2];
            co3 = w[ia3 * 2 + 1];
            si3 = w[ia3 * 2];
            i1 = ia1 + ie;
            for (i0 = 0; i0 < n; i0++) {
                i2 = i1 + n2;
                i3 = i2 + n2;
                r1 = x[2 * i0] + x[2 * i2];
                r2 = x[2 * i0] - x[2 * i2];
                t = x[2 * i1] + x[2 * i3];
                x[2 * i0] = r1 + t;
                r1 = r1 - t;
                s1 = x[2 * i0 + 1] + x[2 * i2 + 1];
                s2 = x[2 * i0 + 1] - x[2 * i2 + 1];
                t = x[2 * i1 + 1] + x[2 * i3 + 1];
                x[2 * i0 + 1] = s1 + t;
                s1 = s1 - t;
                w[2*i2] = (r1 * co2 + s1 * si2) >> 15;
                w[2*i2 + 1] = (s1 * co2 - r1 * si2)>>15;
                t = x[2 * i1 + 1] - x[2 * i3 + 1];
                r1 = r2 + t;
                r2 = r2 - t;
                t = x[2 * i1] - x[2 * i3];
            }
        }
    }
}
```

86 Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences
s1 = s2 - t;
s2 = s2 + t;

x[2 * i1] = (r1 * co1 + s1 * si1) >> 15;
x[2 * i1 + 1] = (s1 * co1 - r1 * si1) >> 15;
x[2 * i3] = (r2 * co3 + s2 * si3) >> 15;
x[2 * i3 + 1] = (s2 * co3 - r2 * si3) >> 15;

DESCRIPTION

This routine is used to compute FFT of a complex sequence of size n, a power of 4, with "decimation-in-frequency decomposition" method. The output is in digit-reversed order. Each complex value is with interleaved 16-bit real and imaginary parts.

TECHNIQUES

1. Loading input x as well as coefficient w in word.
2. Both loops j and i0 shown in the C code are placed in the INNERLOOP of the assembly code.

ASSUMPTIONS

4 <= n <= 65536
Both input x and coefficient w should be aligned on word boundary.

MEMORY NOTE

Align x and w on different word boundaries to minimize memory bank hits. There are N/4 memory bank hits total.

CYCLES

LOGBASE4(N) * (10 * N/4 + 33) + 7 + N/4

*******************************************************************************

.global dummy
.global _radix4
.bss dummy, 52 ; reserve space for dummy
.text
_radix4:

MVK .S1 dummy, A0 ; new dummy pointer in A0 and B1
MVK .S2 dummy, B1

MVKH .S1 dummy, A0 ; new dummy pointer in A0 and B1
MVKH .S2 dummy, B1

STW .D2 B3, *B1 ; push return address on dummy
STW .D1 A10, *+A0[1] ; push A10 on dummy

*** BEGIN Benchmark Timing ***
B_START:

MVK .S1 32, A1 ; A1 = 32
LMBD .L1 1, A4, A2 ; 31 - log2(n)
SHR .S2X A4, 2, B6 ; n2 = n / 4
ZERO .L2 B7 ; mask

SUB .L1 A1, A2, A4 ; log2(n)+1 (circ buff size in bytes)
SHR .S1 A4, 1, A7 ; 2 * n2 = n / 2, a-side
SHR .S2X A4, 1, B9 ; 2 * n2 = n / 2, b-side
MV .L2 B6, B0 ; n / 4
STW .D1 A12, *+A0[5] ; push A12 on dummy

SHL .S1 A4, 16, A4 ; shift into BK0 field
MVC .S2 B4, IRP ; save off x
STW .D1 A13, *+A0[7] ; push A13 on dummy

ADDK .S1 0404h, A4 ; A5, B5 set circular mode on BK0
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

```
MVK .S2 1, B8 ; ie = 1
STW .D1 A14, *+A0[9] ; push A14 on dummy
MVC .S2X A4, AMR ; load AMR
SUB .L2 B0, 1, B0 ; loop counter = n / 4 - 1

K_LOOP:
MV .L2 B4, B5 ; reset X load pointer
MV .L1X B4, A5 ; reset X store pointer
ADD .D2 B0, 1, B1 ; i = loop counter + 1
MV .D1 A6, A1 ; setup twiddle factor pointer
ZERO .S1 A4 ; j = 0
SUBAW .D1 A5, A7, A5 ; setup for first preincrement
AND .S2 B1, B7, B1 ; j loop twiddle reload test
SUBAW .D1 A5, A7, A5 ; setup for first preincrement
MPY .M2 B1, 1, B2 ; j loop twiddle reload test
LDW .D2 *B5+[B6], B10 ; xi0=xt[0*n2], yi0 = yt[0*n2+1]
LDW .D2 *B5+[B6], A8 ; xi1=xt[2*n2], yi1 = yt[2*n2+1]
LDW .D2 *B5+[B6], B11 ; xi2=xt[4*n2], yi2 = yt[4*n2+1]
LDW .D2 *B5+[B6], A9 ; xi3=xt[6*n2], yi3 = yt[6*n2+1]
NOP 2

[[!B2] LDW .D1 *+A1[A4], A13 ; si3 = w[6*j], co3 = w[6*j+1]
[[!B2] ADD .L1X A4, B8, A4 ; j += ie
SUB2 .S2 B10, B11, B3 ; r2a=xi0 - xi2,sa2 = yi0 - yi2
ADD .L2 B0, 0, B1 ;* i = loop counter
MV .L1 A6, A1 ; reset w
SUB2 .S1 A8, A9, A10 ; t3=x1i - x3, t1 = yi1 - yi3
AND .S2 B1, B7, B1 ;* j loop twiddle reload test
ADD2 .S1 A8, A9, A8 ; t0=x1i + x3, t2 = yi1 + yi3
ADD2 .S2 B10, B11, B1 ; r1a=xi0 + xi2,sa1 = yi0 + yi2
MPY .M2 B1, 1, B2 ;* j loop twiddle reload test
[[!B1] ADDAW .D2 B5, 1, B5 ;* reset x input, (circular)
[[!B2] SUBAW .D1 A5, 1, A5 ;
ADD .L2 B0, 1, B0 ;
ZERO .L1 A2 ; first pass cond. init to zero

LOOP:
SHR .S1X B3, 16, A9 ;* extract s2a
SHR .S2X A10, 16, B10 ;* extract t1
[[!B2] ADDAW .D1 A5, 1, A5 ;* reset x output, (circular)
LDW .D2 *B5+[B6], B10 ;** xi0=xt[0*n2], yi0=yt[0*n2+1]
MV .L1 A6, A1 ;** reset w
ADD .L2 B3, B10, B11 ;* r1c = r2a + t1
SUB .L1 A9, A10, A12 ;* s1c = s2a - t3
SUB2 .S2X B1, A8, B1 ;* rlb=rla - t0, s1b = sla - t2
ADD2 .S1X B1, A8, A8 ;* xo0=rla + t0, yo0 = sla + t2
LDW .D2 *B5+[B6], A8 ;** xi0=xt[2*n2], yi0=yt[2*n2+1]
[[!B2] LDW .D1 *+A1[A4], A3 ;** si2 = w[4*j], co2 = w[4*j+1]

[A2] ADD .S2X A11, 2, B3 ; copy B-side x store pointer
[A2] ADD .S2X A11, 2, B3 ; copy B-side x store pointer
[A2] SHR .S1X A14, 15, A14 ; xo2 = xa2 >> 15
SUB .L2 B3, B10, B12 ;* r2c = r2a - t1
ADD .L1 A9, A10, A9 ;* s2c = s2a + t3
MPY .M1X A12, B15, A10 ;* ss1 = s1c * s1i
MPY .M1X A12, B15, A10 ;* ss1 = s1c * s1i
LDW .D2 *B5+[B6], B11 ;** xi2=xt[4*n2], yi2=yt[4*n2+1]
[[!B2] LDW .D1 *+A1[A4], A3 ;** si2 = w[4*j], co2 = w[4*j+1]
[A2] SHR .S2 B13, 15, B13 ; yo1 = ya1 >> 15
[A2] SHR .S1 A15, 15, A15 ; xo3 = xa3 >> 15

Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences
```
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

```
MPYLH .MIX B1, A3, A10 ;* rc2 = slb * co2
MPYLH .M2X B12, B15, B11 ;* sc1 = slc * co1
LDW .D2 *B5+[B6], A9 ;** xi3=xt[6*n2], yi3=yt[6*n2+1]
ADDAW .D1 A5, A7, A5 ;

[A2] SHR .S2 B14, 15, B14 ; yo2 = ya2 >> 15
[B0] B .S1 LOOP ; for i
  MPY .M1 A9, A13, A12 ;* ss3 = s2c * si3
  ADD .L1X B10, A10, A8 ;* xa1 = rc1 + ss1
  MYPY .M2X B11, B13, A13 ;* rs1 = rc1 * si3
  STW .D1 A8, *++A5[A7] ;* xt[0*n2]=xo0, yt[0*n2+1]=yo0

[A2] SHR .S2 B4, 15, B4 ; yo3 = ya3 >> 15
[B2] LDW .D1 A9, A13, A1 ;* ss3 = s2c * si3
MYPY .M2X B1, A3, B11 ;* sc1 = slb * co1
MPYLH .M1X B1, A3, A9 ;* ss2 = slb * si2

MPYLH .M1 A9, A13, A1 ;* sc3 = s2c * co3
MPY .M2X B1, A3, B12 ;* rs2 = rlb * si2
SUB .L2 B11, B13, B13 ;* ya1 = sc1 - rs1
SUB2 .S1 A8, A9, A14 ;* xa2 = rc2 + ss2
SUB2 .S2 B10, B11, B3 ;* rs2 = x10-x12, s2a = y10-yi2
SUB .L2 B0, 1, B0 ;*** generate loop counter

MPYLH .M1X B12, A13, A11 ;* rc3 = r2c * co3
ADD .L1 A10, A9, A14 ;* xa2 = rc2 + ss2
SUB2 .S2 B10, B11, B3 ;* rs2 = x10-x12, s2a = y10-yi2
SUB2 .S2 B14, B12, B14 ;* ya2 = sc2 - rs2

SUB .L2 B14, B12, B14 ;* ya2 = sc2 - rs2
SUB2 .S1 A8, A9, A10 ;** t3=x11 - xi3, t1 = y11-yi3
AND .S2 B1, B7, B1 ;*** j loop twiddle reload test

[A2] ADD .L2 B1, B7, B1 ;*** j = 0 loop counter - 1

[B1] ADDAW .D2 B5, 1, B5 ;*** reset x input, (circular)

; LOOP ends here

SHL .S2 B7, 2, B7 ; mask <<< 2
MVP .M2 B6, B8, B0 ; n/4 = n2 * ie
SHR .S1 A7, 2, A7 ; 2 * n2 >> 2
SHR .S1 B9, 2, B9 ; 2 * n2 >> 2
ADD .L2 B7, 3, B7 ; mask += 3
SHR .S2 B6, 2, B6 ; n2 >> 2
SUB .L2 B0, 1, B0 ; loop counter = n/4 - 1
CMPT .L2 B7, B0, B1 ; kcond = mask > n/4 - 1

[B1] B .S1 K_LOOP ; if (!kcond) do loop
SHL .S2 B8, 2, B8 ; ie <<< 2
MVC .S2 IRP, B4 ; reload x
NOP 4

; K_LOOP ends here

B_END:
*** END Benchmark Timing ***
```

Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences 89
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

MVKH .S1 dummy, A0 ; new dummy pointer in A0 and B0
MVKH .S2 dummy, B0 ; new dummy pointer in A0 and B0

LDW .D2 *B0, B3 ; pop return address off dummy

ZERO .L2 B2

LDW .D2 *+A0[1], A10 ; pop A10 off dummy
LDW .D2 *+B0[2], B10 ; pop B10 off dummy
MVC .S2 B2, AMR ; reset AMR

LDW .D1 *+A0[3], A11 ; pop A11 off dummy
LDW .D2 *+B0[4], B11 ; pop B11 off dummy
LDW .D1 *+A0[5], A12 ; pop A12 off dummy
LDW .D2 *+B0[6], B12 ; pop B12 off dummy
LDW .D1 *+A0[7], A13 ; pop A13 off dummy
LDW .D2 *+B0[8], B13 ; pop B13 off dummy
LDW .D1 *+A0[9], A14 ; pop A14 off dummy
LDW .D2 *+B0[10], B14 ; pop B14 off dummy

B .S2 B3

LDW .D1 *+A0[11], A15 ; pop A15 off dummy
LDW .D2 *+B0[12], B15 ; pop B15 off dummy

NOP 4 ; wait 4 cycles for the last pop
 ; to occur before returning
Example 28. digit.asm File

; FILE
;   digit.asm - C62xx assembly source for a C callable FFT digit reversal
;   function.
;*******************************************************************************
; DESCRIPTION
;   This functions implements, by table look up, digit/bit reversal for FFT
;   algorithms. The function assumes that index tables which contain the
;   indexes of data pairs that get swapped are pre-computed and stored as
;   two separate arrays. Since this is a table look up method, this is a
;   generic routine. It can be used for bit-reversal of radix-2 FFTs, or
;   digit-reversal of radix-4 FFTs etc.
;*******************************************************************************
; PLOTOTYPE
;   void digit_reverse(int *yx, unsigned short *IIndex,
;                      unsigned short *JIndex, int count)
;*******************************************************************************
; IMPLEMENTATION
;   The following C code is functional equivalent to this assembly version.
;*******************************************************************************
; void digit_reverse(int *yx, unsigned short *JIndex,
;                    unsigned short *IIndex, int count)
;{
;    int i;
;    unsigned short I, J;
;    int YXI, YXJ;
;    for (i = 0; i<count; i++)
;    {
;        I = IIndex[i];
;        J = JIndex[i];
;        YXJ = yx[I];
;        YXI = yx[J];
;        yx[J] = YXI;
;        yx[I] = YXJ;
;    }
;}
;*******************************************************************************

.global _digit_reverse

AXPtr .set a4 ; arg1 passed by calling function
; pointer to FFT data, this is a static
; pointer, data to be reversed is accessed
; using indexes, also this an A register,
; thus it is used in the .d1 unit
JIndexPtr .set b4 ; arg2 passed by calling function
; pointer to digit reversal index
IIndexPtr .set a6 ; arg3 passed by calling function
; pointer to other digit reversal index
count .set b6 ; arg4 passed by calling function
; number of points to reverse

J .set a0 ; index loaded using JIndex pointer
I .set b0 ; index loaded using IIndex pointer
TJ .set a7 ; temporary copy of J, this is needed
; because the next value of J is loaded
; before the current one is finished being used
; it is used to store the data value
; loaded by the I index
Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences

```
TI .set b7 ; temporary copy of I, this is needed
      ; because the next value of I is loaded
      ; before the current one is finished being used
      ; it is used to store the data value
      ; loaded by the J index

XI .set a5 ; data value loaded using the I index

XJ .set b5 ; data value loaded using the J index

BXPtr .set b2 ; pointer to FFT data, points to the same
      ; memory location as AXPtr, it is a B register,
      ; so it can be used in the .d2 unit

CNT .set b1 ; count register, used for looping

_digit_reverse:
  ldh .d1 *IIndexPtr++[1], I ; load a I index
  || ldh .d2 *JIndexPtr++[1], J ; load a J index
  mv .l12 AXPtr, BXPtr ; copy AXPtr to BXPtr
  nop 2 ; fill the delay slots
  ldh .d1 *IIndexPtr++[1], I ; load the next I index
  || ldh .d2 *JIndexPtr++[1], J ; load the next J index
  sub .l2 count,1,CNT ; decrement the count by
  || one, and put into a register
  || that can be used as a
  || condition register
  nop 1 ; fill a delay slot
  ldw .d1 *+AXPtr[J],XJ ; load the value pointed by
  || ldw .d2 *+BXPtr[I],XI ; load the value pointed by
  || b LOOP ; branch for the first time
  || through the loop
  nop ; fill a delay slot
  mv .l1 J,TJ ; make a copy of J so that
  || mv .l2 I,TI ; make a copy of I so that
  ||[CNT] b .s1 LOOP ; conditional branch, branch
  ||[!CNT]b .s2 B3 ; having the return
  ; here may be a bug,
  ; we can try it when
  ; we get everything else
  ; working

  ldh .d1 *IIndexPtr++[1], I ; load the next I index
  || ldh .d2 *JIndexPtr++[1], J ; load the next J index
  [CNT] sub .l2 CNT,1,CNT ; decrement the loop counter

  stw .d1 XI,**AXPtr[TJ] ; data loaded from the I index
  || stw .d2 XJ,**BXPtr[TI] ; data loaded from the J index
```

Note: The text appears to be a mix of comments and assembly code. The comments and assembly code are not directly translatable into plain text as they are used in a specific programming language and context. The assembly code snippet seems to be part of an algorithm for implementing the Fast Fourier Transform (FFT) for real-valued sequences.
mv .l1 J, TJ  ; make a copy of J so that
            ; the value is not lost due
            ; to the reloading of J
mv .l2 I, TI  ; make a copy of I so that
            ; the value is not lost due
            ; to the reloading of I

b .s2 B3      ; function return
nop 5
References


